## 4A,5V, 500KHz Synchronous Buck Converter

#### **Features**

- High Efficiency up to 94%
  - Automatic Skip/PWM Mode Operation
- Adjustable Output Voltage from 0.8V to V IN
- Operating from 2.9 to 6V Supply
- Integrated 85m& High Side / 75m& Low Side MOSFETs
- Low Dropout Operation: 100% Duty Cycle
- Stable with Low ESR Ceramic Capacitors
- Current up to 4A
- Power-On-Reset Detection on VCC and VIN
- Integrate Soft-Start and Soft-Stop
- Over-Temperature Protection
- Over-Voltage Protection
- Under-Voltage Protection
- High/ Low Side Current Limit
- Power Good Indication
- Enable/Shutdown Function
- Available in SOP-8P and TDFN3x3-10 Packages
- Lead Free and Green Devices Available (RoHS Compliant)

### **Applications**

- Notebook Computer & UMPC
- LCD Monitor/TV
- Set-Top Box
- DSL, Switch HUBr
- Portable Instrument

### **General Description**

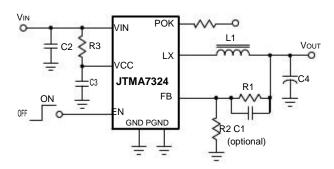
JTMA7324 is a synchronous buck converters with integrated 85m & high side and 75m & low side power MOSFETs. The JTMA7324 with a current-mode control scheme can convert wide input voltage of 2.9V to 6V to the output voltage adjustable from 0.8V to 6V to provide excellent output voltage regulation.

The JTMA7324 is equipped with an automatic Skip/PWM mode operation. At light load , the IC operates in the Skip mode to reduce the switching losses. At heavy load, the IC works in PWM mode.

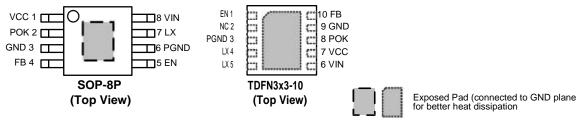
The JTMA7324 is also equipped with Power-on-reset, soft start, soft-stop, and whole protections (under-voltage, over-voltage, over-temperature and current-limit) into a single package.

This device, available in SOP-8P and TDFN3x3-10 packages, provides a very compact system solution external components and PCB area.

### **Simplified Application Circuit**

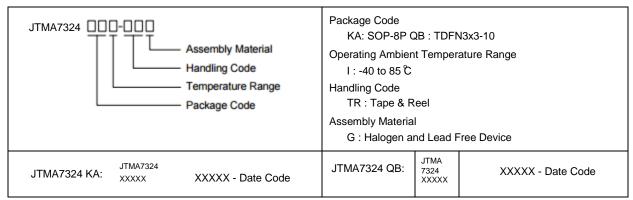


## **Pin Configuration**



JIATAIMU reserves the right to make changes to improve reliability or manufacturability without notice, and advise customers to obtain the latest version of relevant information to verify before placing orders.

### **Ordering and Marking Information**



Note: JIATAIMU lead-free products contain molding compounds/die attach materials and 100% matte tin plate termination finish; which are fully compliant with RoHS. JIATAIMU lead-free products meet or exceed the lead-free requirements of IPC/JEDEC J-STD-020D for MSL classification at lead-free peak reflow temperature. JIATAIMU defines "Green" to mean lead-free (RoHS compliant) and halogen free (Br or Cl does not exceed 900ppm by weight in homogeneous material and total of Br and Cl does not exceed 1500ppm by weight).

### **Absolute Maximum Ratings (Note 1)**

Symbol	Parameter	Rating	Unit
VIN, VCC	VIN and VCC Input Voltage	-0.3 ~ 7	V
VLX	LX to GND Voltage	-1 ~Vcc+0.3	V
	FB, EN, POK to GND Voltage	-0.3 ~ 6.5	V
Po	P <sub>D</sub> Power Dissipation		W
TJ	Junction Temperature	150	°C
Тѕтс	Storage Temperature	-65 ~ 150	°C
Tsdr	T <sub>SDR</sub> Maximum Lead Soldering Temperature(10 Seconds)		°C

Note1: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability

#### **Thermal Characteristics**

Symbol	Parameter	Typical Value	Unit
\JA	Junction-to-Ambient Resistance in Free Air (Note 2)	50	°C/W
(JC	Junction-to-Case Resistance in Free Air (Note 3)	10	°C/W

Note 2:  $\bigcup_{A}$  is measured with the component mounted on a high effective thermal conductivity test board in free air. The exposed pad of SOP-8P is soldered directly on the PCB.

Note 3: The case temperature is measured at the center of the exposed pad on the underside of the SOP-8P package.

## **Recommended Operating Conditions** (Note 4)

Symbol	Parameter	Range	Unit
Vcc	VCC Supply Voltage	2.9~ 6	V
Vin	VIN Supply Voltage	2.5~6	V
Vоит	Converter Output Voltage	0.8~6	V
Іоит	Converter Maximum Output Current	4	Α
TA	Ambient Temperature	-40 ~ 85	°C
TJ	Junction Temperature	-40 ~ 125	°C

Note 4: Refer to the typical application circuit.

### **Electrical Characteristics**

Unless otherwise specified, these specifications apply over  $V_{\text{CC}}=V_{\text{IN}}=5V$ ,  $V_{\text{OUT}}=3.3V$  and  $T_{\text{A}}=-40\sim85$  °C. Typical values are at  $T_{\text{A}}=25$  °C.

Symbol	Parameter	Test Conditions	APW7324			Unit	
Syllibol	raiailletei	rest Conditions	Min.	Тур.	Max.		
SUPPLY	CURRENT	'	•	•		1	
Ivcc	VCC Supply Current	V <sub>FB</sub> =0.7V	-	460	-	∞A	
Ivcc_sd	VCC Shutdown Supply Current	EN=GND	-	-	20	∞A	
POWER-	ON-RESET (POR)						
	VCC POR Voltage Threshold	Vcc Rising	=	2.7	-	V	
	VCC POR Voltage Hysteresis	Vcc Falling	=	200	-	mV	
	VCC Debounce Time		-	10	-	∞s	
	VIN POR Voltage Threshold	V <sub>IN</sub> Rising	-	2.3	-	V	
	VIN POR Voltage Hysteresis	V <sub>IN</sub> Falling	-	50	-	mV	
REFERE	NCE VOLTAGE	'	•				
VREF	Reference Voltage		-	0.8	-	V	
VKEF	Thereferice voltage	All temperature	-1	-	+1	%	
	Output Accuracy	Іоит=10mA~3A, Vcc=2.9~5V	-1.5	-	+1.5	%	
OSCILLA	TOR and DUTY CYCLE	'	•				
Fosc	Oscillator Frequency		425	500	575	KHz	
	Maximum Converter's Duty	V <sub>FB</sub> =0.7V	-	100	-	%	
	Minimum on Time		-	70	-	ns	
POWER	MOSFET	•			1	ı	
	High Side MOSFET Resistance	Vcc=Vin=5V, ILx=0.5A, Ta=25°C	-	85	115	m&	
	Low Side MOSFET Resistance	Vcc=Vin=5V, ILx=0.5A, Ta=25°C	-	75	105	m&	

### **JTMA7324**

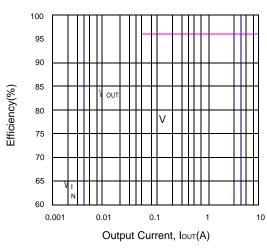
### **Electrical Characteristics (Cont.)**

Unless otherwise specified, these specifications apply over  $V_{CC}=V_{IN}=5V$ ,  $V_{OUT}=3.3V$  and  $T_A=-40\sim85$  C. Typical values are at  $T_A=25$  C.

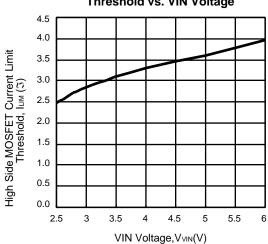
Symbo	Parameter	Test Conditions		APW7324		Unit
Syllibo	Farameter	rest Conditions	Min.	Тур.	Max.	Unit
CURREN	T -MODE PWM CONVERTER			· ·	•	
Gm	Error Amplifier Transconductance		-	680	-	∞A/V
	Error Amplifier DC Gain	COMP=NC	-	50	-	dB
	Current Sense Resistance		-	400	-	m&
T□	Dead Time		-	20	-	ns
PROTEC	EIONS					·
Ішм	High Side MOSFET current-limit	Peak Current	5	6	7	Α
Тотр	Over-temperature Trip Point		-	160	-	°C
	Over-temperature Hysteresis		-	50	-	°C
	Over- Voltage Protection Threshold		120	125	130	%V <sub>REF</sub>
	OVP Debounce Time		-	3	-	∝s
	Under-Voltage Protection Threshold		45	50	55	%V <sub>REF</sub>
	UVP Debounce Time		-	3	-	∞s
	Low Side Switch Current-Limit	From Drain to Source	-1.6	-1.9	-2.2	Α
SOFT-ST	ART, ENABLE and INPUT CURRENT	rs	I			I.
	Soft Start Time		-	2.5	-	ms
	EN Enable Threshold Voltage	VEN Rising	0.8	-	1.3	V
	EN Threshold Hysteresis		-	0.15	-	V
	EN Pull High Resistance		-	300	-	K&
POWER (	GOOD		I			I.
		POK in from lower (POK goes high)	85	87.5	90	%Vouт
	POK Threshold	POK low hysteresis	-	5	i	%Vоит
		POK in from higher (POK goes high)	110	112.5	115	%Vоит
		POK high hysteresis	-	5	-	%Vоит
	POK Pull Low Resistance		-	1	1	⊇&
	POK Low-to-High Debounce Time		-	0.5	-	ms

## **Typical Operating Characteristics**

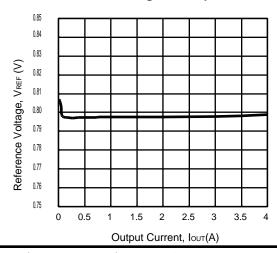




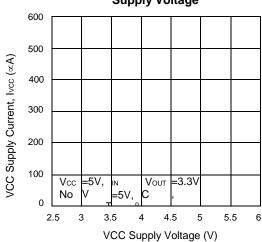
### High Side MOSFET Current Limit Threshold vs. VIN Voltage



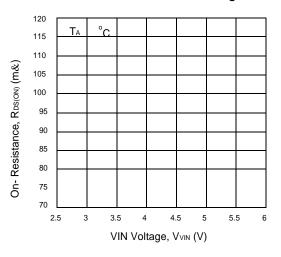
#### Reference Voltage vs. Output Current



# VCC Supply Current vs. VCC Supply Voltage



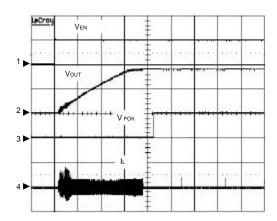
#### On-Resistance vs. VIN Voltage



### **Operating Waveforms**

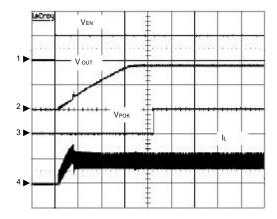
The test condition is JTMA7324,  $V_{IN}=V_{CC}=5V$ ,  $V_{OUT}=1.8V$ ,  $C_{OUT}=22 \times Fx2$ ,  $L=2.2 \times H$ ,  $T_{A}=25 \cdot C$  unless otherwise specified.

#### **Enable without Loading**



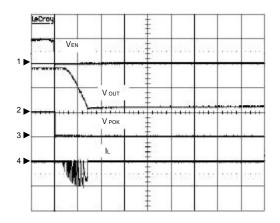
No load, EN Power On CH1: V<sub>EN</sub>, 5V/Div, DC CH2: V<sub>OUT</sub>, 1V/Div, DC CH3: V<sub>POK</sub>, 5V/Div, DC CH4: I<sub>L</sub>, 2A/Div, DC TIME: 1ms/Div

#### **Enable with 2A Loading**



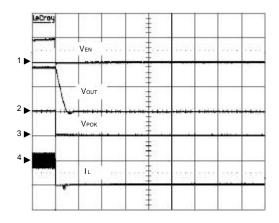
IOUT=2A, EN Power On CH1: VEN, 5V/Div, DC CH2: VOUT, 1V/Div, DC CH3: VPOK, 5V/Div, DC CH4: IL, 2A/Div, DC TIME: 1ms/Div

#### **Shutdown without Loading**



No load, EN Power Off CH1: V<sub>EN</sub>, 5V/Div, DC CH2: V<sub>OUT</sub>, 1V/Div, DC CH3: V<sub>POK</sub>, 5V/Div, DC CH4: I<sub>L</sub>, 2A/Div, DC TIME: 100∝s/Div

#### Shutdown with 2A Loading

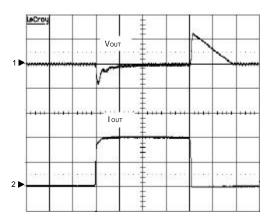


IOUT=2A, EN Power Off CH1: VEN, 5V/Div, DC CH2: VOUT, 1V/Div, DC CH3: VPOK, 5V/Div, DC CH4: IL, 2A/Div, DC TIME: 100∝S/Div

### **Operating Waveforms**

The test condition is JTMA7324, VIN=Vcc=5V, Vout=1.8V, Cout =22 xFx2, L=2.2 xH, TA= 25 C unless otherwise specified.

#### **Load Transient**

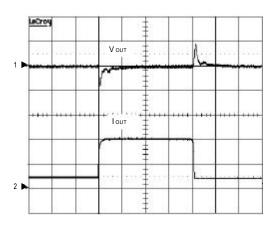


Iouт=100mA-4A-100mA, Slew Rate=1A/∞s

CH1: Vout, 200mV/Div, offset=1.8V

CH2: Iout, 2A/Div, DC TIME: 50∝s/Div

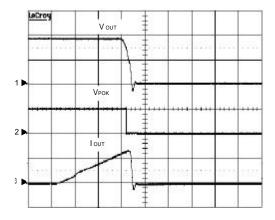
#### **Load Transient**



Iou $\tau$ =1A-4A-1A, Slew Rate=1A/ $\infty$ s CH1: Vou $\tau$ , 200mV/Div, offset=1.8V

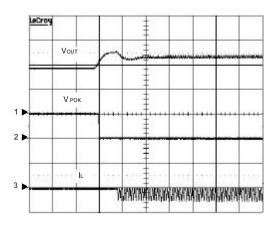
CH2: Iout, 2A/Div, DC TIME: 50∝s/Div

#### **Current Limit**



CH1: Vout ,1V/Div, DC CH2: VPOK, 5V/Div, DC CH3: Iout, 5A/Div, DC TIME: 200∞s/Div

#### **Over Voltage Protection**



No Load, a 2.5V power source is externally

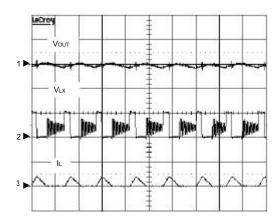
attached to Vout

CH1: Vout, 1V/Div, DC CH2: VPOK, 5V/Div, DC CH3: IL, 2A/Div, DC TIME: 50\inftysepssylves

### **Operating Waveforms**

The test condition is JTMA7324,  $V_{IN}=V_{CC}=5V$ ,  $V_{OUT}=1.8V$ ,  $C_{OUT}=22 \times Fx2$ ,  $L=2.2 \times H$ ,  $T_{A}=25 \cdot C$  unless otherwise specified.

#### **Normal Operation in Ligh Load**

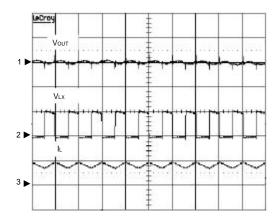


Iout=200mA

CH1: Vout, 50mV/Div, offset=1.8V

CH2: V<sub>L</sub>x, 5V/Div, DC CH3: I<sub>L</sub>x, 2A/Div, DC TIME: 2∞s/Div

#### **Normal Operation in Heavy Load**



Iout=4A

CH1: Vout, 50mV/Div, offset=1.8V

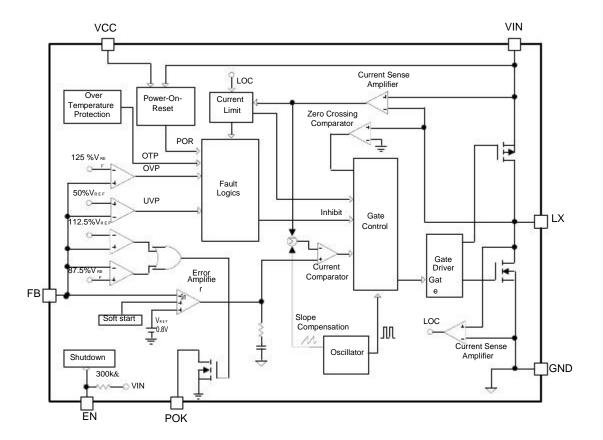
CH2: V<sub>L</sub>x, 5V/Div, DC CH3: I<sub>L</sub>x, 5A/Div, DC TIME: 2∞s/Div

## **JTMA7324**

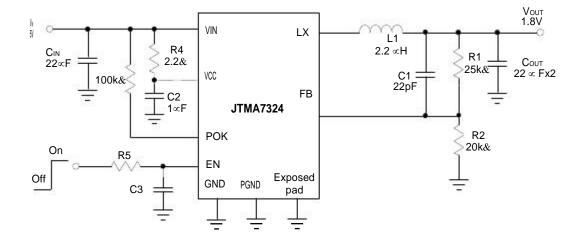
# **Pin Description**

	Pin		Function	
NO.		Name	i unction	
SOP-8P	TDFN3x3-10			
1	7	VCC	Signal Input. VCC supplies the control circuitry, gate drivers. Connecting a ceramic bypass capacitor from VCC to GND to eliminate switching noise and voltage ripple on the input to	
2	8	POK	Power Good Output. This pin is open-drain logic output that is pulled to ground when the output voltage is not within ±12.5% of regulation point. The POK pin, if used, needs an external pull high resistor in the range of 30k&~100k&	
3	9	GND	Ground. Power and signal ground.	
4	10	FB	Output Feedback Input. The APW7324 senses the feedback voltage via FB and regulates the voltage at 0.8V. Connecting FB with a resistor-divider from the converter's output sets the output voltage.	
5	1	EN	Enable Input. EN is a digital input that turns the regulator on or off. Driving the EN high turns on the regulator, otherwise, driving it low turns it off. The EN pin is internally pulled high via a $300 k\Omega$ resistor.	
6	3	PGND	Power Ground.	
7	4, 5	LX	Power Switching Output. LX is the Junction of the high-side and low-side Power MOSFETs to supply power to the output LC filter.	
8	8 6 VIN F		Power Input. VIN supplies the step-down converter switches.	
9	11	Exposed Pad	Exposed pad. Connect the exposed pad to the system ground plane with large copper area for dissipating heat into the ambient air.	

## **Block Diagram**



# **Typical Application Circuit**



### **Function Description**

#### VCC and VIN Power-On-Reset (POR)

TThe JTMA7324 keeps monitoring the voltage on VIN and VCC pins to prevent wrong logic operations which may occur when VIN or VCC voltage is not high enough for internal control circuitry to operate. The VCC POR rising threshold is 2.7V (typical) with 0.2V hysteresis and VIN POR rising threshold is 2.3V with 0.05V hysteresis. During startup, the VCC and VIN voltage must exceed the POR threshold. Then the IC starts a starts-up process and ramps up the output voltage to the voltage target.

#### **Output Under-Voltage Protection (UVP)**

In the operational process, if a short-circuit occurs, the output voltage will drop quickly. Before the current-limit circuit responds, the output voltage will fall out of the required regulation range. The under-voltage continually monitors the FB voltage after soft-start is completed. If a load step is strong enough to pull the output voltage lower than the under-voltage threshold, the IC shuts down converter's output.

The under-voltage threshold is 50% of the nominal output voltage. The under-voltage comparator has a built-in  $3 \times s$  noise filter to prevent the chips from wrong UVP shutdown being caused by noise. The JTMA7324 will be latched after under-voltage protection.

#### **Over-Voltage Protection (OVP)**

The over-voltage function monitors the output voltage by FB pin. When the FB voltage increases over 125% of the reference voltage due to the high-side MOSFET failure or for other reasons, the over-voltage protection comparator will force the low-side MOSFET gate driver high. This action actively pulls down the output voltage and eventually attempts to blow the internal bonding wires. As soon as the output voltage is within regulation, the OVP comparator is disengaged. The chip will restore its normal operation.

#### **Over-Temperature Protection (OTP)**

The over-temperature circuit limits the junction temperature of the JTMA7324. When the junction temperature exceeds  $T_J = 160 \, {}_{\circ}\text{C}$ , a thermal sensor turns off the both power MOSFETs, allowing the devices to cool.

The thermal sensor allows the converters to start a startup process and to regulate the output voltage again after the junction temperature cools by 50°C. The OTP is designed with a 50 °C hysteresis to lower the average TJ during continuous thermal overload conditions, increasing lifetime of the JTMA7324.

#### **Current-Limit Protection**

The JTMA7324 monitors the output current, flows through the high-side and low-side power MOSFETs, and limits the current peak at current-limit level to prevent the IC from damaging during overload, short-circuit and overvoltage conditions. Typical high side power MOSFET current limit is 6A, and low side MOSFET current limit is 1.9A.

#### Soft-Start

The JTMA7324 has a built-in soft-start to control the rise rate of the output voltage and limit the input current surge during start-up. During soft-start, an internal voltage ramp connected to one of the positive inputs of the error amplifier, rises up from 0V to 0.95V to replace the reference voltage (0.8V) until the voltage ramp reaches the reference voltage. During soft-start without output overvoltage, the JTMA7324 converter's sinking capability is disabled until the output voltage reaches the voltage target.

#### Soft-Stop

At the moment of shutdown controlled by EN signal, under-voltage event or over-temperature protection, the JTMA7324 initiates a soft-stop process to discharge the output voltage in the output capacitors. Certainly, the load current also discharges the output voltage. During soft-stop, the internal voltage ramp (VRAMP) falls down from 0. 95V to 0V to replace the reference voltage. Therefore, the output voltage falls down slowly at the light load. After the soft-stop interval elapses, the soft-stop process ends and the IC turns on the low-side power MOSFET.

#### **Enable and Shutdown**

Driving EN to ground places the JTMA7324 in shutdown. In shutdown mode, the internal power MOSFETs turns off, all internal circuitry shuts down and the quiescent supply current reduces to less than  $20 \times A$ .

### **JTMA7324**

### **Function Description (Cont.)**

#### **Power Good Indicator**

POK is actively held low in shutdown and soft-start status. In the soft-start process, the POK is an open-drain. When the soft-start is finished, the POK is released. In normal operation, the POK window is from 87.5% to 112.5% of the converter reference voltage. When the output voltage has to stay within this window, POK signal will become high after 0.5ms internal delay. When the output voltage outruns 82.55% or 117.5% of the target voltage, POK signal will be pulled low immediately. In order to prevent false POK drop, capacitors need to parallel at the output to confine the voltage deviation with severe load step transient.

### **Application Information**

#### **Input Capacitor Selection**

Because buck converters have a pulsating input current, a low ESR input capacitor is required. This results in the best input voltage filtering, minimizing the interference with other circuits caused by high input voltage spikes. Also, the input capacitor must be sufficiently large to stabilize the input voltage during heavy load transients. For good input voltage filtering, usually a  $22 \times F$  input capacitor is sufficient. It can be increased without any limit for better input voltage filtering. Ceramic capacitors show better performance because of the low ESR value, and they are less sensitive against voltage transients and spikes compared to tantalum capacitors. Place the input capacitor as close as possible to the input and GND pin of the device for better performance.

#### **Inductor Selection**

For high efficiencies, the inductor should have a low dc resistance to minimize conduction losses. Especially at high-switching frequencies the core material has a higher impact on efficiency. When using small chip inductors, the efficiency is reduced mainly due to higher inductor core losses. This needs to be considered when selecting the appropriate inductor. The inductor value determines the inductor ripple current. The larger the inductor value, the smaller the inductor ripple current and the lower the conduction losses of the converter. Conversely, larger inductor values cause a slower load transient response.

A reasonable starting point for setting ripple current,  $\Box l_L$  is 40% of maximum output current. The recommended inductor value can be calculated as below:

 $I_{L(MAX)} = I_{OUT(MAX)} + 1/2 \times \square I_{L}$ 

To avoid saturation of the inductor, the inductor should be rated at least for the maximum output current of the converter plus the inductor ripple current.

#### **Output Voltage Setting**

In the adjustable version, the output voltage is set by a resistive divider. The external resistive divider is connected to the output, allowing remote voltage sensing as

shown in "Typical Application Circuits". A suggestion of maximum value of R2 is 300k& to keep the minimum current that provides enough noise rejection ability through the resistor divider. The output voltage can be calculated as below:

$$Vout = VREF \oplus \sqrt[|]{1 + \frac{R!}{R!}} \mid Vout$$

$$R1\delta1M\&$$

$$FB$$

$$JTMA7324$$

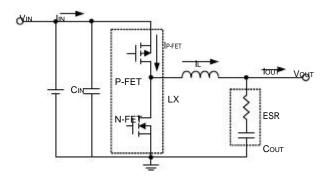
$$GND$$

$$R2 \delta 300K\&$$

#### **Output Capacitor Selection**

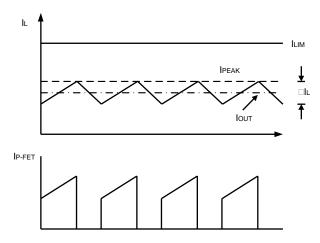
The current-mode control scheme of the JTMA7324 allows the use of tiny ceramic capacitors. The higher capacitor value provides the good load transients response. Ceramic capacitors with low ESR values have the lowest output voltage ripple and are recommended. If required, tantalum capacitors may be used as well. The output ripple is the sum of the voltages across the ESR and the ideal output capacitor.

When choosing the input and output ceramic capacitors, choose the X5R or X7R dielectric formulations. These dielectrics have the best temperature and voltage characteristics of all the ceramics for a given value and size.



### **Application Information (Cont.)**

#### **Output Capacitor Selection (Cont.)**

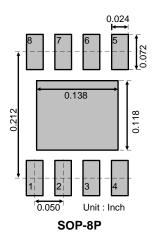


## Layout Considerations

For all switching power supplies, the layout is an important step in the design; especially at high peak currents and switching frequencies. If the layout is not carefully done, the regulator might show noise problems and duty cycle jitter.

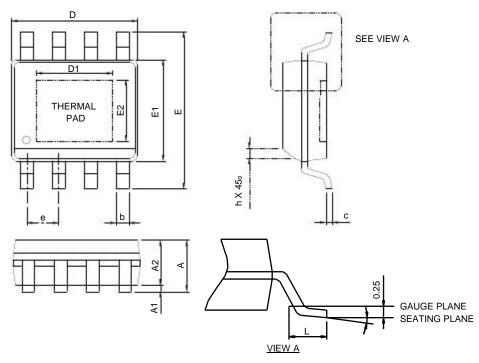
- 1. The input capacitor should be placed close to the VIN and GND. Connect the capacitor and VIN/GND with short and wide trace without any via holes for good input voltage filtering. The distance between VIN/GND to capacitor less than 2mm respectively is recommended.
- To minimize copper trace connections that can inject noise into the system, the inductor should be placed as close as possible to the LX pin to minimize the noise coupling into other circuits.
- 3. The output capacitor should be place closed to VOUT and GND.
- 4. Keep the sensitive small signal nodes (FB) away from switching nodes (LX) on the PCB. Therefore place the feedback divider and the feedback compensation network close to the IC to avoid switching noise. Connect the ground of feedback divider directly to the GND pin of the IC using a dedicated ground trace.
- 5. A star ground connection or ground plane minimizes ground shifts and noise is recommended.

#### **Recommended Minimum Footprint**



## **Package Information**

#### SOP-8P



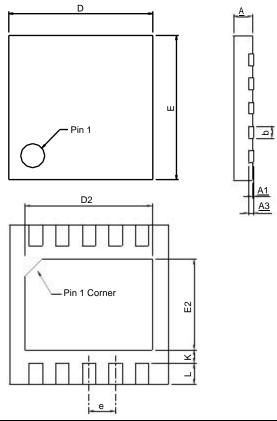
S		SOP-	8P	
S Y M B O L	MILLIME	TERS	INCHE	S
Ö	MIN.	MAX.	MIN.	MAX.
Α		1.60		0.063
A1	0.00	0.15	0.000	0.006
A2	1.25		0.049	
b	0.31	0.51	0.012	0.020
С	0.17	0.25	0.007	0.010
D	4.80	5.00	0.189	0.197
D1	2.50	3.50	0.098	0.138
E	5.80	6.20	0.228	0.244
E1	3.80	4.00	0.150	0.157
E2	2.00	3.00	0.079	0.118
е	1.27 E	SSC	0.050 E	BSC
h	0.25	0.50	0.010	0.020
L	0.40	1.27	0.016	0.050
0	೨೦	28	20	28

Note: 1. Followed from JEDEC MS-012 BA.

- 2. Dimension "D" does not include mold flash, protrusions or gate burrs. Mold flash, protrusion or gate burrs shall not exceed 6 mil per side .
- Dimension "E" does not include inter-lead flash or protrusions. Inter-lead flash and protrusions shall not exceed 10 mil per side.

# **Package Information**

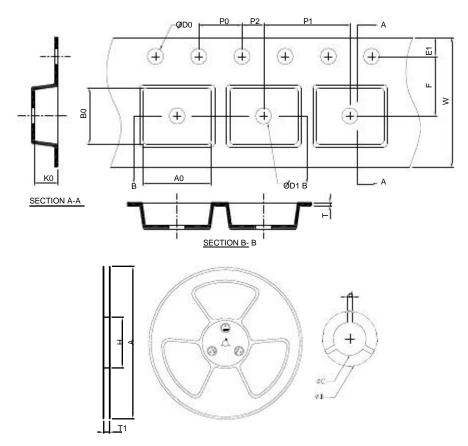
TDFN3x3-10



S		TDFN3	x3-10	
SYMBO.	MILLIME	TERS	INCI	HES
Ď	MIN.	MAX.	MIN.	MAX.
Α	0.70	0.80	0.028	0.031
A1	0.00	0.05	0.000	0.002
А3	0.20	REF	0.008	REF
b	0.18	0.30	0.007	0.012
D	2.90	3.10	0.114	0.122
D2	2.20	2.70	0.087	0.106
Е	2.90	3.10	0.114	0.122
E2	1.40	1.75	0.055	0.069
е	0.50	BSC	0.020	BSC
L	0.30	0.50	0.012	0.020
K	0.20		0.008	

Note: 1. Followed from JEDEC MO-229 VEED-5.

# **Carrier Tape & Reel Dimensions**



Application	Α	Н	T1	С	d	D	W	E1	F
SOP-8P	330.0±2.00	50 MIN.	12.4+2.00 -0.00	13.0+0.50 -0.20	1.5 MIN.	20.2 MIN.	12.0±0.30	1.75±0.10	5.5±0.05
SUP-8P	P0	P1	P2	D0	D1	Т	A0	В0	K0
	4.0±0.10	8.0±0.10	2.0±0.05	1.5+0.10 -0.00	1.5 MIN.	0.6+0.00 -0.40	6.40±0.20	5.20±0.20	2.10±0.20
Application	Α	Н	T1	С	d	D	W	E1	F
TDENSV2 40	330.0±2.00	50 MIN.	12.4 <b>+</b> 2.00 -0.00	13.0+0.50 -0.20	1.5 MIN.	20.2 MIN.	12.0±0.30	1.75±0.10	5.5±0.05
TDFN3x3-10	P0	P1	P2	D0	D1	T	A0	В0	K0
	4.0±0.10	8.0±0.10	2.0±0.05	1.5+0.10 -0.00	1.5 MIN.	0.6+0.00 -0.40	3.30±0.20	3.30±0.20	1.30±0.20

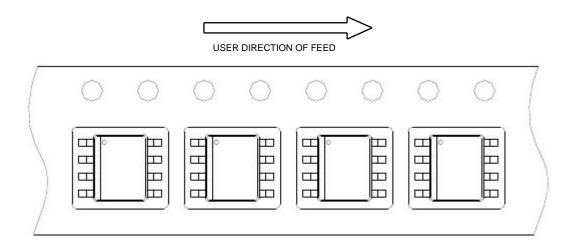
(mm)

### **Devices Per Unit**

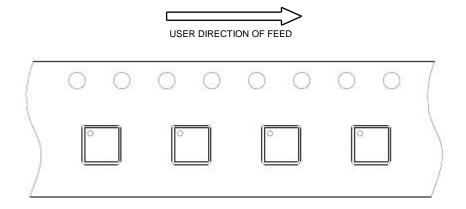
Package Type	Unit	Quantity
SOP- 8P	Tape & Reel	2500
TDFN-3x3-10	Tape & Reel	3000

# **Taping Direction Information**

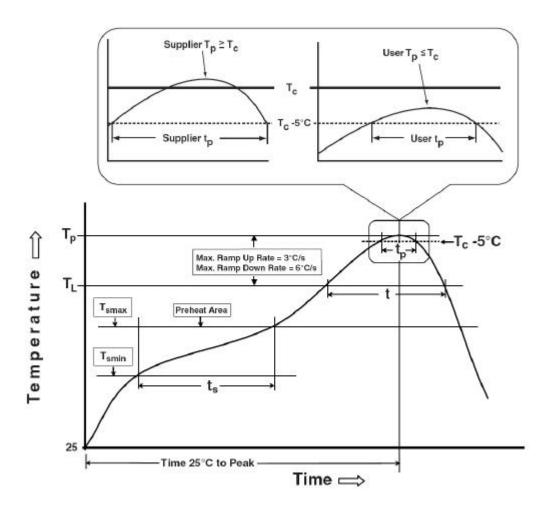
SOP-8P



#### TDFN3x3-10



### **Classification Profile**



### **Classification Reflow Profiles**

Sn-Pb Eutectic Assembly	Pb-Free Assembly
100 °C 150 °C 60-120 seconds	150 °C 200 °C 60-120 seconds
3 °C/second max.	3°C/second max.
183 °C 60-150 seconds	217 °C 60-150 seconds
See Classification Temp in table 1	See Classification Temp in table 2
20** seconds	30** seconds
6 °C/second max.	6 °C/second max.
6 minutes max.	8 minutes max.
	100 °C 150 °C 60-120 seconds  3 °C/second max.  183 °C 60-150 seconds  See Classification Temp in table 1  20** seconds  6 °C/second max.

<sup>\*</sup> Tolerance for peak profile Temperature  $(T_{\text{P}})$  is defined as a supplier minimum and a user maximum.

Table 1. SnPb Eutectic Process – Classification Temperatures (Tc)

Package Thickness	Volume mm <sup>3</sup> <350	Volume mm <sup>3</sup> ε350
<2.5 mm	235 °C	220 °C
ε2.5 mm	220 °C	220 °C

Table 2. Pb-free Process – Classification Temperatures (Tc)

Package Thickness	Volume mm <sup>3</sup> <350	Volume mm <sup>3</sup> 350-2000	Volume mm <sup>3</sup> >2000
<1.6 mm	260 °C	260 °C	260 °C
1.6 mm – 2.5 mm	260 °C	250 °C	245 °C
ε2.5 mm	250 °C	245 °C	245 °C

## **Reliability Test Program**

Test item	Method	Description	
SOLDERABILITY	JESD-22, B102	5 Sec, 245°C	
HOLT	JESD-22, A108	1000 Hrs, Bias @ Tj=125°C	
PCT	JESD-22, A102	168 Hrs, 100%RH, 2atm, 121°C	
тст	JESD-22, A104	500 Cycles, -65°C~150°C	
НВМ	MIL-STD-883-3015.7	VHBM≧2KV	
MM	JESD-22, A115	VMM≧200V	
Latch-Up	JESD 78	10ms, 1tr≥100mA	

<sup>\*\*</sup> Tolerance for time at peak profile temperature (tp) is defined as a supplier minimum and a user maximum.

# JTMA7324

## **Customer Service**