ETR28009-001a

50mA/200mA Inductor Built-in Step-Down "micro DC/DC" Converters

☆GreenOperationCompatible

■GENERAL DESCRIPTION

The JTMT210 series is a synchronous step-down micro DC/DC converter which integrates an inductor and a control IC in one tiny package (2.0mm×2.5mm, h=1.0mm). An internal coil simplifies the circuit and enables minimization of noise and other operational trouble due to the circuit wiring. A wide operating voltage range of 2.0V to 6.0V enables support for applications that require an internally fixed output voltage from 1.0V to 4.0V in increments of 0.05V.

During stand-by, all circuits are shutdown to reduce currentconsumption to as low as 0.1 µA or less.

With the built-in UVLO (Under Voltage Lock Out) function, the internal P-channel MOS driver transistor is forced OFF when input voltage becomes UVLO ditect Voltage or lower.

The JTMT210 integrate C_L discharge function which enables the electric charge at the output capacitor C_L to be discharged via the internal discharge switch located between the Lx and Vss pins. When the devices enter stand-by mode, output voltage quickly returns to the Vss level as a result of this function.

■APPLICATIONS

- Wearable Devices
- Smart meters
- Bluetooth units
- Energy Harvest devices
- Back up power supply circuits
- Portable game consoles
- Devices with 1 Lithium cell

■FEATURES

Input Voltage : 2.0V~6.0V

Output Voltage : 1.0V~4.0V (±2.0%, 0.05V step increments)

Control Methods : PFM control

Output Current : 200mA(JTMT210A/JTMT210C)

50mA(JTMT210B/JTMT210D)

Supply Current : 0.5µA

High Efficiency : 93% (V_{IN}=3.6V,VouT=3.0V/ 100μ A)

Function : UVLO

Short Circuit Protection

C_L Discharge

Capacitor : Low ESR Ceramic Capacitor

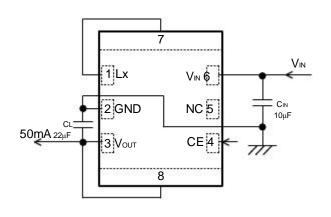
Operating Ambient Temperature : $-40^{\circ}\text{C} \sim +85^{\circ}\text{C}$ Packages : CL-2025-02

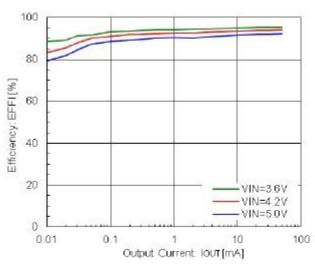
Environmentally Friendly : EU RoHS Compliant, Pb Free

■TYPICAL APPLICATION CIRCUIT

■ TYPICAL PERFORMANCE CHARACTERISTICS

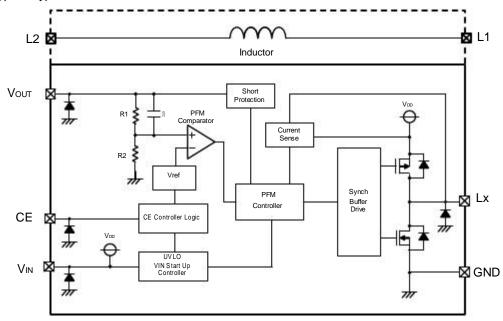
JTMT210B301GR-G(Vout=3.0V)





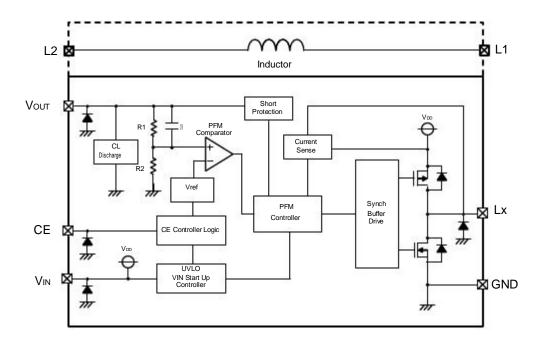
■BLOCK DIAGRAM

JTMT210 Series, Type A/TypeB



- * JTMT210A and B type do not have $C_{\text{\tiny L}}$ Discharge function.
- * Diodes inside the circuits are ESD protection diodes and parasitic diodes.

JTMT210 Series, Type C/TypeD



■PRODUCT CLASSIFICATION

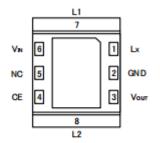
Ordering information

JTMT210123456-7

DESIGNATOR	ITEM	SYMBOL	DESCRIPTION
		А	Іо∪т=200mA , Without C∟ Auto Discharge
1	Product Type	В	Іо∪т=50mA Without C∟ Auto Discharge
		С	louт=200mA , With C∟ Auto Discharge
		D	Іо∪т=50mA, With C∟ Auto Discharge
23	Output Voltage	10~40	Output voltage options e.g.) 1.2V → ② =1 ③ =2 1.25V→ ② =1 ③ =C 0.05V increments: 0.05=A, 0.15=B, 0.25=C, 0.35=D, 0.45=E, 0.55=F, 0.65=H, 0.75=K, 0.85=L, 0.95=M
4)	Fixed number	1	Fixed number
(5)(6)-(7) ^(*1)	Package (Order Unit)	GR-G	CL-2025-02 (3,000pcs/Reel)

The "-G" suffix denotes Halogen and Antimony free as well as being fully EU RoHS compliant.

■PIN CONFIGURATION



* The dissipation pad for the CL-2025-02 package should be solder-plated in recommended mount pattern and metal masking so as to enhance mounting strength and heat release.

The mount pattern should be connected to GND pin (No.2).

(BOTTOM VIEW)

■PIN ASSIGNMENT

PIN NUMBER	PIN NAME	FUNCTIONS
1	Lx	Switching
2	GND	Ground
3	Vouт	Output Voltage
4	CE	Chip Enable
5	NC	Ground
6	Vin	Power Input
7	L1	Inductor Electrodes
8	L2	Inductor Electrodes

■CE PIN FUNCTION

PIN NAME	SIGNAL	STATUS
CE	Н	Operation (All Types)
	L	Standby (All Types)

^{*} Please do not leave the CE pin open.

■ABSOLUTE MAXIMUM RATINGS

Ta=25°C

PARAMETER	SYMBOL	RATINGS	UNITS
V _{IN} Pin Voltage	Vin	-0.3 ~ +7.0	٧
Lx Pin Voltage	VLX	-0.3 ~ V _{IN} +0.3 or +7.0 (*1)	V
Vout Pin Voltage	Vouт	-0.3 ~ V _{IN} +0.3 or +7.0 (*1)	V
CE Pin Voltage	Vce	-0.3 ~ +7.0	٧
Lx Pin Current	lıx	1000	mA
Power Dissipation	Pd	1000 (*2)	mW
Operating Ambient Temperature	Topr	-40 ~ +85	°C
Storage Temperature	Tstg	-55 ~ +125	°C

^{*} All voltages are described based on the GND.

Please refer to page 15 for details.

The maximum value is the lower of either V_{IN} + 0.3 or +7.0.

 $^{^{1}}$ The power dissipation figure shown is PCB mounted (40mm×40mm, t=1.6mm, Glass Epoxy FR-4).

■ELECTRICAL CHARACTERISTICS

●JTMT210Axx1GR-G, without CL discharge function

Ta=25°C

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Input Voltage	Vin	-	2.0	-	6.0	V	1)
Output Voltage	V _{OUT(E)} (*2)	Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while Vout is decreasing.		E1		V	2
UVLO Release Voltage	Vuvlo(e)	$V_{CE}=V_{IN},\ V_{OUT}=0V.$ Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while V_{IN} is increasing.	1.65	1.80	1.95	V	2
UVLO Hysteresis Voltage	V _{HYS(E)}	$V_{CE}=V_{IN},\ V_{OUT}=0V.\ Resistor\ connected\ with\ Lx\ pin.$ $V_{UVLO(E)}\ Voltage\ which\ Lx\ pin\ changes\ "H"\ to\ "L"}$ level while V_{IN} is decreasing.	0.11	0.15	0.24	V	2
Supply Current	lq	$V_{\text{IN=Vce=Vout(T)}} + 0.5V$, $V_{\text{IN=2}}^{(1)} = 2.0V$, if $V_{\text{out(T)}} \le 1.5V$ $V_{\text{IN=Vce=Vout(T)}} + 0.5V$, $V_{\text{IN=Open}}^{(1)} = 0.5V$		E2		μA	3
Standby Current	Іѕтв	V _{IN} =5.0V, V _{CE} =V _{OUT} =0V, L _X =Open.	-	0.1	1.0	μA	3
Lx SW "H" Leak Current	ILEAKH	Vin=5.0V, Vce=Vout=0V, Vlx=0V.	-	0.1	1.0	μA	3
Lx SW "L" Leak Current	ILEAKL	VIN=5.0V, VCE=VOUT=0V, VLX=5.0V.	-	0.1	1.0	μA	3
PFM Switching Current	Ірғм	Vin=Vce=Vout(t)+2.0V , lout=10mA.	260	330	400	mA	1)
Maximum Duty Ratio (*3)	MAXDTY	V _{IN} =V _{OUT} =V _{OUT} (T)x0.95V , V _{CE} =1.2V Resistor connected with L _x pin.	100	-	-	%	2
Efficiency (*4)	EFFI	VIN=VCE=5.0V, VOUT(T)=4.0V, IOUT=30mA.	-	93	-	%	6
Efficiency (*4)	EFFI	VIN=VCE=3.6V, VOUT(T)=3.3V, IOUT=30mA.	-	93	-	%	6
Efficiency (*4)	EFFI	V _{IN} =V _{CE} =3.6V, V _{OUT(T)} =1.8V, I _{OUT} =30mA.	-	87	-	%	6
Lx SW "Pch" ON Resistance (*5)	R _{LXP}	VIN=VCE=5.0V, VOUT=0V, ILX=100mA.	-	0.4	0.65	Ω	4
Lx SW "Nch" ON Resistance	RLXN	VIN=VCE=5.0V.	-	0.4 (*6)	-	Ω	-
Output Voltage Temperature Characteristics	ΔVουτ/ (Vουτ • ΔTopr)	-40°C≦Topr≦85°C.	-	±100	-	ppm/℃	2
CE "High" Voltage	V _{СЕН}	Vout=0V. Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while V_{CE} =0.2 \rightarrow 1.5V.	1.2	-	6.0	V	5
CE "Low" Voltage	VCEL	Vout=0V. Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while V_{CE} =1.5 \rightarrow 0.2V.	GND	-	0.3	V	\$
CE "High" Current	СЕН	VIN=VcE=5.0V, Vout=0V, Lx=Open.	-0.1	-	0.1	μA	5
CE "Low" Current	ICEL	V _{IN} =5.0V, V _{CE} =V _{OUT} =0V, L _X =Open.	-0.1	-	0.1	μA	5
Short Protection Threshold Voltage	Vshort	Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while $V_{OUT} = V_{OUT(I)} + 0.1V \rightarrow 0V$.	0.4	0.5	0.6	V	2
Inductance Value	L	Test Frequency=1MHz	-	8.0	-	μH	
(Coil) Rated Current	IDC_L	ΔT=+40°C	-	600	-	mA	

Unless otherwise stated, V_{IN} = V_{CE} =5.0V

The actual output voltage value $V_{\text{OUT}(E)}$ is the PFM comparator threshold voltage in the IC.

Therefore, the DC/DC circuit output voltage, including the peripheral components, is boosted by the ripple voltage average value.

 $^{(^{\}circ}1)$ Vout(T)=Nominal Output Voltage

 $^{(^{\}circ}2)$ $V_{OUT(E)}$ =Effective Output Voltage

 $^{^{(^{\}prime 3})}$ Not applicable to the products with $V_{\text{OUT}(T)}$ < 2.15V since it is out of operational volatge range.

⁽⁴⁾ EFFI=[{ (Output Voltage)x(Output Current)] / [(Input Voltage)x(Input Current)}]x100

 $^{^{(^{1}5)}}$ Lx SW "Pch" ON resistance = (V $_{IN} - V_{Lx}$ pin measurement voltage) / 100mA

^(*6) Designed value

●JTMT210Bxx1GR-G, without CL discharge function

Ta=25°℃

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Input Voltage	Vin	-	2.0	-	6.0	V	1
Output Voltage	V _{OUT(E)} (*2)	Resistor connected with Lx pin.Voltage which Lx pin changes "L" to "H" level while Vout is decreasing.		E1		V	2
UVLO Release Voltage	Vuvlo(e)	$V_{\text{CE}} = V_{\text{IN}}, V_{\text{OUT}} = 0V.$ Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while V_{IN} is increasing.	1.65	1.80	1.95	V	2
UVLO Hysteresis Voltage	V _{HYS(E)}	$\begin{split} &V_{CE} \! = \! V_{IN}, V_{OUT} \! = \! 0V. \text{Resistor connected with Lx pin.} \\ &V_{UVLO(E)} \! - Voltage \text{which Lx pin changes "H" to "L"} \\ &\text{level while } V_{IN} \text{is decreasing.} \end{split}$	0.11	0.15	0.24	V	2
Supply Current	lq	$V_{\text{IN}}=V_{\text{CE}}=V_{\text{OUT}(T)}+0.5V$, $V_{\text{IN}}^{(T)}=2.0V$, if $V_{\text{OUT}(T)} \leq 1.5V$ $V_{\text{IN}}=V_{\text{CE}}=V_{\text{OUT}(T)}+0.5V$, $V_{\text{IN}}=V_{\text{CE}}=V_{\text{OUT}(T)}+0.5V$		E2		μA	3
Standby Current	Іѕтв	V _{IN} =5.0V, V _{CE} =V _{OUT} =0V, L _X =Open.	-	0.1	1.0	μA	3
Lx SW "H" Leak Current	ILEAKH	VIN=5.0V, VCE=VOUT=0V, VLX=0V.	-	0.1	1.0	μА	3
Lx SW "L" Leak Current	ILEAKL	VIN=5.0V, VCE=VOUT=0V, VLX=5.0V.	-	0.1	1.0	μA	3
PFM Switching Current	РЕМ	V _{IN} =V _{CE} =V _{OUT(T)} +2.0V , l _{OUT} =10mA.	115	180	250	mA	1)
Maximum Duty Ratio (*3)	MAXDTY	V _{IN} =V _{OUT} =V _{OUT(T)} x0.95V , V _{CE} =1.2V Resistor connected with L _x pin.	100	-	-	%	2
Efficiency (*4)	EFFI	VIN=VCE=5.0V, VOUT (T)=4.0V, IOUT=30mA.	-	95	-	%	6
Efficiency (*4)	EFFI	VIN=VCE=3.6V, VOUT(T)=3.3V, IOUT=30mA.	-	95	-	%	6
Efficiency (*4)	EFFI	VIN=VCE=3.6V, VOUT(T)=1.8V, IouT=30mA.	-	89	-	%	6
Lx SW "Pch" ON Resistance	R _{LXP}	VIN=VCE=5.0V, VOUT=0V, ILX=100mA.	-	0.4	0.65	Ω	4
Lx SW "Nch" ON Resistance	RLXN	Vin=Vce=5.0V.	-	0.4 (*6)	-	Ω	-
Output Voltage Temperature Characteristics	ΔVουτ/ (Vουτ • ΔTopr)	-40°C≦Topr≦85°C.	-	±100	-	ppm/℃	2
CE "High" Voltage	V _{СЕН}	Vout=0V. Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while V_{CE} =0.2 \rightarrow 1.5V.	1.2	-	6.0	V	5
CE "Low" Voltage	Vcel	Vout=0V. Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while V_{CE} =1.5 \rightarrow 0.2V.	GND	-	0.3	V	\$
CE "High" Current	Ісен	V _{IN} =V _{CE} =5.0V, V _{OUT} =0V, L _X =Open.	-0.1	-	0.1	μA	5
CE "Low" Current	ICEL	V _{IN} =5.0V, V _{CE} =V _{OUT} =0V, Lx=Open.	-0.1	-	0.1	μA	(5)
Short Protection Threshold Voltage	Vshort	Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while Vouт=Vouт(r)+0.1V→0V . ("1)	0.4	0.5	0.6	V	2
Inductance Value	L	Test Frequency=1MHz	-	8.0	-	μH	
(Coil) Rated Current	loc_L	ΔT=+40℃	-	600	-	mA	

Unless otherwise stated, V_{IN} = V_{CE} =5.0V

The actual output voltage value $V_{\text{OUT}(E)}$ is the PFM comparator threshold voltage in the IC.

Therefore, the DC/DC circuit output voltage, including the peripheral components, is boosted by the ripple voltage average value.

 $^{(^{\}circ}1)$ Vout(T)=Nominal Output Voltage

 $^{(^{\}circ}2)$ $V_{OUT(E)}$ =Effective Output Voltage

 $^{^{(^{\}circ}\!3)}$ Not applicable to the products with $V_{OUT(T)}\!<2.15V$ since it is out of operational volatge range.

^(*4) EFFI=[{ (Output Voltage)x(Output Current)] / [(Input Voltage)x(Input Current)}]x100

 $^{^{(^{1}5)}}$ Lx SW "Pch" ON resistance = (V $_{IN} - V_{LX} \, pin$ measurement voltage) / 100mA

^(*6) Designed value

●JTMT210Cxx1GR-G, with CL Discharge Function

Ta=25°C

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Input Voltage	Vin	-	2.0	-	6.0	V	1)
Output Voltage	V _{OUT(E)} (*2)	Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while Vout is decreasing.		E1			2
UVLO Release Voltage	Vuvlo(e)	V _{CE} =V _{IN} , V _{OUT} =0V. Resistor connected with L _x pin. Voltage which L _x pin changes "L" to "H" level while V _{IN} is increasing.	1.65	1.80	1.95	V	2
UVLO Hysteresis Voltage	V _{HYS(E)}	V _{CE} =V _{IN} , V _{OUT} =0V. Resistor connected with L _x pin. V _{UV} LO(E) - Voltage which L _x pin changes "H" to "L" level while V _{IN} is decreasing.	0.11	0.15	0.24	V	2
Supply Current	lq	$V_{\text{IN}}=V_{\text{CE}}=V_{\text{OUT}(T)}+0.5V$ $V_{\text{IN}}^{(1)}=2.0V$, if $V_{\text{OUT}(T)} \le 1.5V$. $V_{\text{OUT}}=V_{\text{OUT}(T)}+0.5V$, $V_{\text{LX}}^{(1)}=Open$.		E2		μА	3
Standby Current	Іѕтв	V _{IN} =5.0V, V _{CE} =V _{OUT} =0V, L _X =Open.	-	0.1	1.0	μA	3
Lx SW "H" Leak Current	ILEAKH	VIN=5.0V, VCE=VOUT=0V, VLX=0V.	-	0.1	1.0	μA	3
Lx SW "L" Leak Current	ILEAKL	VIN=5.0V, VCE=VOUT=0V, VLX=5.0V.	-	0.1	1.0	μА	3
PFM Switching Current	РЕМ	V _{IN} =V _{CE} =V _{OUT(T)} +2.0V , I _{OUT} =10mA.	260	330	400	mA	1)
Maximum Duty Ratio (*3)	MAXDTY	$V_{\text{IN}}=V_{\text{OUT}}=V_{\text{OUT}(T)}$ x0.95V , $V_{\text{CE}}^{(\tau)}$ 1.2V Resistor connected with Lx pin.	100	-	-	%	2
Efficiency (*4)	EFFI	V _{IN} =V _{CE} =5.0V, V _{OUT(T)} =4.0V , lour=30mA.	-	93	-	%	6
Efficiency (*4)	EFFI	VIN=VCE=3.6V, VOUT(T)=3.3V , IouT=30mA.	-	93	-	%	6
Efficiency (*4)	EFFI	VIN=VCE=3.6V, VOUT(T)=1.8V, lout=30mA.	-	87	-	%	6
L _x SW "Pch" ON Resistance (*5)	RLXP	V _{IN} =V _{CE} =5.0V, V _{OUT} =0V, I _{LX} =100mA.	-	0.4	0.65	Ω	4
Lx SW "Nch" ON Resistance	RLXN	VIN=VCE=5.0V.	-	0.4 (*6)	-	Ω	-
Output Voltage Temperature Characteristics	ΔVουτ/ (Vουτ • ΔTopr)	-40°C≦Topr≦85°C.	-	±100	-	ppm/℃	2
CE "High" Voltage	Vсен	Vout=0V. Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while V_{CE} =0.2 \rightarrow 1.5V.	1.2	-	6.0	V	(5)
CE "Low" Voltage	Vcel	Vour=0V. Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while Vc∈=1.5→0.2V.	GND	-	0.3	V	(5)
CE "High" Current	Ісен	V _{IN} =V _{CE} =5.0V, V _{OUT} =0V, L _X =Open.	-0.1	-	0.1	μA	(5)
CE "Low" Current	ICEL	V _{IN} =5.0V, V _{CE} =V _{OUT} =0V, L _X =Open.	-0.1	-	0.1	μA	(5)
Short Protection Threshold Voltage	Vshort	Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while Vouт= Vouт(r)+0.1V→0V . ("1)	0.4	0.5	0.6	V	2
C _L Discharge	Rdchg	V _{IN} =V _{OUT} =5.0V, V _{CE} =0V, L _X =Open.	55	80	105	Ω	3
Inductance Value	L	Test Frequency=1MHz	-	8.0	-	μH	
(Coil) Rated Current	IDC_L	ΔT=+40°C	-	600	-	mA	

Unless otherwise stated, V_{IN} = V_{CE} =5.0V

Therefore, the DC/DC circuit output voltage, including the peripheral components, is boosted by the ripple voltage average value.

^(*1) Vout(t)=Nominal Output Voltage

 $^(^{*}2)$ $V_{OUT(E)}$ =Effective Output Voltage

The actual output voltage value $V_{\text{OUT}(E)}$ is the PFM comparator threshold voltage in the IC.

 $^{^{(3)}}$ Not applicable to the products with $V_{OUT(T)}$ < 2.15V since it is out of operational volatge range.

 $[\]label{eq:continuous} \begin{tabular}{ll} EFFI=[\{\ (Output\ Voltage) \times (Output\ Current)\}]/\ [(Input\ Voltage) \times (Input\ Current)\}] \times 100 \end{tabular}$

 $_{(5)}$ Lx SW "Pch" ON resistance = $(V_{IN} - V_{LX} pin measurement voltage) / 100mA$

Designed value

●JTMT210Dxx1GR-G, with CL Discharge Function

Ta=25°C

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Input Voltage	Vin	-	2.0	-	6.0	V	1)
Output Voltage	Vout(E)	Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while Vour is decreasing.	E1			V	2
UVLO Release Voltage	V _{UVLO(E)}	$V_{\text{CE}}=V_{\text{IN}}, V_{\text{OUT}}=0V$. Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while V_{IN} is increasing.	1.65	1.80	1.95	V	2
UVLO Hysteresis Voltage	V _{HYS(E)}	$V_{CE} = V_{IN}, \ V_{OUT} = 0V. \ Resistor \ connected \ with \ L_x \ pin.$ $V_{UVLO(E)} - Voltage \ which \ L_x \ pin \ changes \ "H" \ to \ "L"$ level while V_{IN} is decreasing.	0.11	0.15	0.24	V	2
Supply Current	lq	$V_{\text{IN}=V_{\text{CE}}=V_{\text{OUT}(T)}}+0.5V$, (*1) $V_{\text{IN}=2.0V}$, if $V_{\text{OUT}(T)} \le 1.5V$, (*1) $V_{\text{OUT}=V_{\text{OUT}(T)}}+0.5V$, $L_{\text{X}}^{(\tau_1)}=\text{Open}$.		E2		μА	3
Standby Current	Іѕтв	VIN=5.0V, VCE=VOUT=0V, Lx=Open.	-	0.1	1.0	μA	3
Lx SW "H" Leak Current	ILEAKH	Vin=5.0V, Vce=Vout=0V, VLx=0V.	-	0.1	1.0	μA	3
Lx SW "L" Leak Current	ILEAKL	Vin=5.0V, Vce=Vout=0V, VLx=5.0V.	-	0.1	1.0	μA	3
PFM Switching Current	Ірғм	V _{IN} =V _{CE} =V _{OUT(T)} +2.0V , I _{OUT} =10mA.	115	180	250	mA	1
Maximum Duty Ratio (*3)	MAXDTY	V _{IN} =V _{OUT} =V _{OUT} (τ)x0.95V , V _{CE} =1.2V Resistor connected with Lx pin.	100	-	-	%	2
Efficiency (*4)	EFFI	VIN=VCE=5.0V, VOUT(T)=4.0V, IOUT=30mA.	-	95	-	%	6
Efficiency (*4)	EFFI	VIN=VCE=3.6V, VOUT(T)=3.3V, IOUT=30mA.	-	95	-	%	6
Efficiency (*4)	EFFI	VIN=VCE=3.6V, VOUT(T)=1.8V , IOUT=30mA.	-	89	-	%	6
L _x SW "Pch" ON Resistance (*5)	R _{LXP}	Vin=Vce=5.0V, Vout=0V, ILX=100mA.	-	0.4	0.65	Ω	4
Lx SW "Nch" ON Resistance	RLXN	VIN=VCE=5.0V.	-	0.4 (*6)	-	Ω	-
Output Voltage Temperature Characteristics	ΔVουτ/ (Vουτ • ΔTopr)	-40°C≦Topr≦85°C.	-	±100	-	ppm/℃	2
CE "High" Voltage	Vсен	V_{OUT} =0V. Resistor connected with Lx pin. Voltage which Lx pin changes "L" to "H" level while V_{CE} =0.2 \rightarrow 1.5V.	1.2	-	6.0	V	(5)
CE "Low" Voltage	Vcel	Vout=0V. Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while V_{CE} =1.5 \rightarrow 0.2V.	GND	-	0.3	V	(5)
CE "High" Current	Ісен	V _{IN} =V _{CE} =5.0V, V _{OUT} =0V, L _X =Open.	-0.1	-	0.1	μA	(5)
CE "Low" Current	ICEL	V _{IN} =5.0V, V _{CE} =V _{OUT} =0V, L _X =Open.	-0.1	-	0.1	μΑ	5
Short Protection Threshold Voltage	Vshort	Resistor connected with Lx pin. Voltage which Lx pin changes "H" to "L" level while Vour= Vour(r)+0.1V→0V. ("1)		0.5	0.6	V	2
C∟ Discharge	Rdchg	V _{IN} =V _{OUT} =5.0V, V _{CE} =0V, L _X =Open.	55	80	105	Ω	3
Inductance Value	L	Test Frequency=1MHz	-	8.0	-	μH	
Rated Current	loc	ΔT=+40°C	-	600	-	mA	

Unless otherwise stated, V_{IN}=V_{CE}=5.0V

The actual output voltage value $V_{\text{OUT}(E)}$ is the PFM comparator threshold voltage in the IC.

Therefore, the DC/DC circuit output voltage, including the peripheral components, is boosted by the ripple voltage average value.

^(*1) Vout(t)=Nominal Output Voltage

^(*2) Vout(E)=Effective Output Voltage

 $^{^{(^3)}}$ Not applicable to the products with $\dot{V}_{\text{OUT(T)}}$ < 2.15V since it is out of operational volatge range.

EFFI=[{ (Output Voltage)x(Output Current)] / [(Input Voltage)x(Input Current)}]x100

 $_{(^5)}$ Lx SW "Pch" ON resistance = (V $_{\text{IN}} - V_{\text{Lx}}$ pin measurement voltage) / 100mA

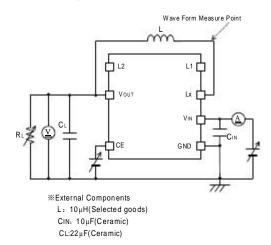
Designed value

JTMT210 Series voltage chart

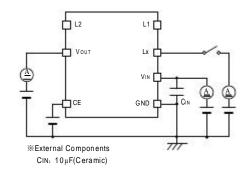
SYMBOL	E	1	E	2	SYMBOL	E	1	E	2
PARAMETER	OUTPUT V	OLTAGE	SUPPLY C	URRENT	PARAMETER	OUTPUT V	OUTPUT VOLTAGE		URRENT
UNITS: V	UNIT	UNITS: V		S: μA	UNITS: V	UNIT	S: V	UNIT	S: µA
OUTPUT	MINI	1407	TVD	MAN	OUTPUT	NAIN.	BAAN/	TVD	MAN
VOLTAGE	MIN.	MAX.	TYP.	MAX.	VOLTAGE	MIN.	MAX.	TYP.	MAX.
1.00	0.980	1.020			3.00	2.940	3.060		
1.05	1.029	1.071			3.05	2.989	3.111		
1.10	1.078	1.122			3.10	3.038	3.162		
1.15	1.127	1.173			3.15	3.087	3.213		
1.20	1.176	1.224	0.500	0.000	3.20	3.136	3.264	0.000	0.400
1.25	1.225	1.275	0.500	0.800	3.25	3.185	3.315	0.800	2.100
1.30	1.274	1.326			3.30	3.234	3.366		
1.35	1.323	1.377	1		3.35	3.283	3.417		
1.40	1.372	1.428	[3.40	3.332	3.468		
1.45	1.421	1.479			3.45	3.381	3.519		
1.50	1.470	1.530			3.50	3.430	3.570		
1.55	1.519	1.581			3.55	3.479	3.621		
1.60	1.568	1.632			3.60	3.528	3.672	1.500	
1.65	1.617	1.683			3.65	3.577	3.723		
1.70	1.666	1.734			3.70	3.626	3.774		
1.75	1.715	1.785	0.500	0.500 0.900	3.75	3.675	3.825		3.000
1.80	1.764	1.836			3.80	3.724	3.876		
1.85	1.813	1.887			3.85	3.773	3.927		
1.90	1.862	1.938			3.90	3.822	3.978		
1.95	1.911	1.989			3.95	3.871	4.029		
2.00	1.960	2.040			4.00	3.920	4.080		
2.05	2.009	2.091							55
2.10	2.058	2.142							
2.15	2.107	2.193							
2.20	2.156	2.244							
2.25	2.205	2.295	0.600	1.100					
2.30	2.254	2.346							
2.35	2.303	2.397							
2.40	2.352	2.448							
2.45	2.401	2.499							
2.50	2.450	2.550			1				
2.55	2.499	2.601							
2.60	2.548	2.652							
2.65	2.597	2.703							
2.70	2.646	2.754	0.700	4.500					
2.75	2.695	2.805	0.700	1.500					
2.80	2.744	2.856							
2.85	2.793	2.907							
2.90	2.842	2.958							
2.30									

■TEST CIRCUITS

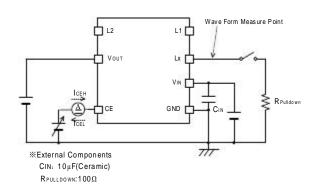
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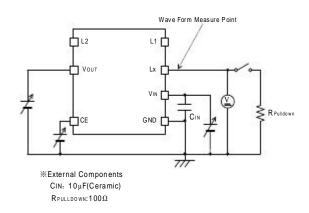
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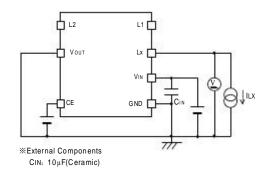
< Circuit No. 5 >



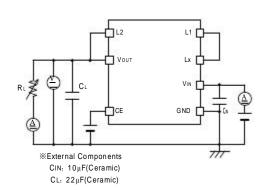
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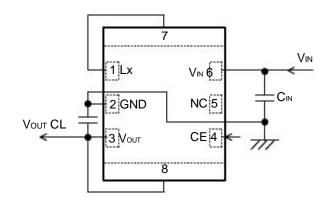
< Circuit No.4 >



< Circuit No.6 >



■TYPICAL APPLICATION CIRCUIT



NOTE:

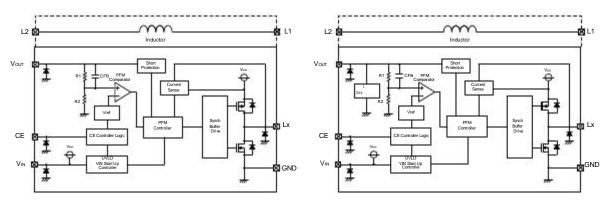
The integrated Inductor can be used only for this DC/DC converter. Please do not use this inductor for other reasons.

	Manufacturer	Part Number	VALUE
	Taiyo Yuden	LMK107BBJ106MALT	10μF/10V
Cin	raiyo ruuen	LMK212ABJ106MG	10μF/10V
	TDK	C1608X5R1A106M	10μF/10V
	TER	C2012X5R1A106M	10μF/10V
	Taiyo Yuden	LMK107BBJ226MA	22μF/10V
CL	raiyo radar	LMK212BBJ226MG	22μF/10V
	TDK	C1608X5R1A226M	22μF/10V
	, bit	C2012X5R1A226M	22µF/10V

 $[\]star$ Take capacitance loss, withstand voltage, and other conditions into consideration when selecting components.

■ OPERATIONAL EXPLANATION

The JTMT210 series consists of a reference voltage supply, PFM comparator, Pch driver Tr, Nch synchronous rectification switch Tr, current sensing circuit, PFM control circuit, CE control circuit, and others. (Refer to the block diagram below.)

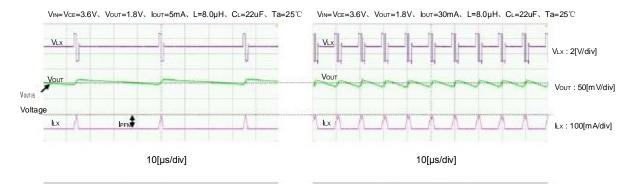


<BLOCK DIAGRAM TYPE A/B>

<BLOCK DIAGRAM TYPE C/D >

An ultra-low quiescent current circuit and synchronous rectification enable a significant reduction of dissipation in the IC, and the IC operates with high efficiency at both light loads and heavy loads. Current limit PFM is used for the control method, and even when switching current superposition occurs, increases of output voltage ripple are suppressed, allowing use over a wide voltage and current range. The IC is compatible with low-capacitance ceramic capacitors, and a small, high-performance step-down DC-DC converter can be created.

The actual output voltage Vour(E) in the electrical characteristics is the threshold voltage of the PFM comparator in the block diagram. Therefore the average output voltage of the step-down circuit, including peripheral components, depends on the ripple voltage. Before use, test fully using the actual device.



<Reference voltage supply (VREF)>

Reference voltage for stabilization of the output voltage of the IC.

<PFM control>

- (1) The feedback voltage (FB voltage) is the voltage that results from dividing the output voltage with the IC internal dividing resistors R_{FB1} and R_{FB2} . The PFM comparator compares this FB voltage to V_{REF} . When the FB voltage is lower than V_{REF} , the PFM comparator sends a signal to the buffer driver through the PFM control circuit to turn on the Pch driver Tr. When the FB voltage is higher than V_{REF} , the PFM comparator sends a signal to prevent the Pch driver Tr from turning on.
- (2) When the Pch driver Tr is on, the current sense circuit monitors the current that flows through the Pch driver Tr connected to the Lx pin. When the current reaches the set PFM switching current (IPFM), the current sense circuit sends a signal to the buffer driver through the PFM control circuit. This signal turns off the Pch driver Tr and turns on the Nch synchronous rectification switch Tr.
- (3) The on time (off time) of the Nch synchronous rectification switch Tr is dynamically optimized inside the IC. After the off time elapses and the PFM comparator detects that the V_{OUT} voltage is higher than the set voltage, the PFM comparator sends a signal to the PFM control circuit that prevents the Pch driver Tr from turning on. However, if the V_{OUT} voltage is lower than the set voltage, the PFM comparator starts Pch driver Tr on.

■OPERATIONAL EXPLANATION (Continued)

By continuously adjusting the interval of the linked operation of (1), (2) and (3) above in response to the load current, the output voltage is stabilized with high efficiency from light loads to heavy loads.

<PFM Switching Current >

The PFM switching current monitors the current that flows through the Pch driver Tr, and is a value that limits the Pch driver Tr current. The Pch driver Tr remains on until the coil current reaches the PFM switching current (IPFM). An approximate value for this on-time ton can be calculated using the following equation:

 $ton = L \times I_{PFM} / (V_{IN} - V_{OUT})$

<Maximum on-time function>

To avoid excessive ripple voltage in the event that the coil current does not reach the PFM switching current within a certain interval even though the Pch driver Tr has turned on and the FB voltage is above VREF, the Pch driver Tr can be turned off at any timing using the maximum on-time function of the PFM control circuit. If the Pch driver Tr turns off by the maximum on-time function instead of the current sense circuit, the Nch synchronous rectification switch Tr will not turn on and the coil current will flow to the Vout pin by means of the parasite diode of the Nch synchronous rectification switch Tr.

<Through mode>

When the V_{IN} voltage is lower than the output voltage, through mode automatically activates and the Pch driver Tr stays on continuously.

- (1) In through mode, when the load current is increased and the current that flows through the Pch driver Tr reaches a load current that is several tens of mA lower than the set PFM switching current (IPFM), the current sense circuit sends a signal through the PFM control circuit to the buffer driver. This signal turns off the Pch driver Tr and turns on the Nch synchronous rectification switch Tr.
- (2) After the on-time (off-time) of the Nch synchronous rectification switch Tr, the Pch driver Tr tums on until the current reaches the set PFM switching current (IPEM) again.

If the load current is large as described above, operations (1) and (2) above are repeated. If the load current is several tens of mA lower than the PFM switching current (I_{PFM}), the Pch driver Tr stays on continuously.

<V_{IN} start mode>

When the V_{IN} voltage rises, V_{IN} start mode stops the short-circuit protection function during the interval until the FB voltage approaches V_{REF}. After the V_{IN} voltage rises and the FB voltage approaches V_{REF} by step-down operation, V_{IN} start mode is released. In order to prevent an excessive rush current while V_{IN} start mode is activated, the coil current flows to the V_{OUT} pin by means of the parasitic diode of the Nch synchronous rectification Tr. In V_{IN} start mode as well, the coil current is limited by the PFM switching current.

<Short-circuit protection function>

The short-circuit protection function monitors the Vout voltage. In the event that the Vout pin is accidentally shorted to GND or an excessive load current causes the Vout voltage to drop below the set short-circuit protection voltage, the short-circuit protection function activates, and tums off and latches the Pch driver Tr at any selected timing. Once in the latched state, the IC is turned off and then restarted from the CE pin, or operation is started by re-applying the V_{IN} voltage.

<UVLO function>

When the V_{IN} pin voltage drops below the UVLO detection voltage, the IC stops switching operation at any selected timing, turns off the Pch driver Tr and Nch synchronous rectification switch Tr (UVLO mode). When the V_{IN} pin voltage recovers and rises above the UVLO release voltage, the IC restarts operation.

<C∟ discharge function>

On the JTMT210 series, a C_L discharge function is available as an option (JTMT210C/JTMT210D types). This function enables quick discharging of the C_L load capacitance when "L" voltage is input into the CE pin by the Nch Tr connected between the V_{OUT} -GND pins, or in UVLO mode. This prevents malfunctioning of the application in the event that a charge remains on C_L when the IC is stopped. The discharge time is determined by C_L and the C_L discharge resistance R_{DCHG} , including the Nch Tr (refer to the diagram below). Using this time constant $\tau = C_L \times R_{DCHG}$, the discharge time of the output voltage is calculated by means of the equation below.

 $V = V_{OUT} \times e_{-t/\tau}$, or in terms of t, $t = \tau ln(V_{OUT} / V)$

V: Output voltage after discharge

Vout : Set output voltage

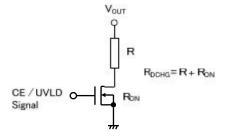
t : Discharge time

C_L: Value of load capacitance (C_L)

RDCHG: Value of CL discharge resistance Varies by power supply voltage.

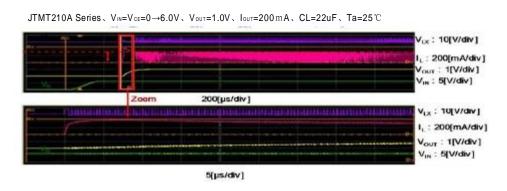
τ: $C_L \times R_{DCHG}$

The C_L discharge function is not available on the JTMT210A/JTMT210B types.



■NOTE ON USE

- 1. Be careful not to exceed the absolute maximum ratings for externally connected components and this IC.
- 2. The DC/DC converter characteristics greatly depend not only on the characteristics of this IC but also on those of externally connected components, so refer to the specifications of each component and be careful when selecting the components. Be especially careful of the characteristics of the capacitor used for the load capacity C_L and use a capacitor with B characteristics (JIS Standard) or an X7R/X5R (EIA Standard) ceramic capacitor.
- 3. Use a ground wire of sufficient strength. Ground potential fluctuation caused by the ground current during switching could cause the IC operation to become unstable, so reinforce the area around the GND pin of the IC in particular.
- 4. Mount the externally connected components in the vicinity of the IC. Also use short, thick wires to reduce the wire impedance.
- 5. When the voltage difference between V_{IN} and V_{OUT} is small, switching energy increases and there is a possibility that the ripple voltage will be too large. Before use, test fully using the actual device.
- 6. The CE pin does not have an internal pull-up or pull-down, etc. Apply the prescribed voltage to the CE pin.
- 7. If other than the recommended inductance and capacitance values are used, excessive ripple voltage or a drop in efficiency may result.
- 8. If other than the recommended inductance and capacitance values are used, a drop in output voltage when the load is excessive may cause the short-circuit protection function to activate. Before use, test fully using the actual device.
- 9. At high temperature, excessive ripple voltage may occur and cause a drop in output voltage and efficiency. Before using at high temperature, test fully using the actual device
- 10. At light loads or when IC operation is stopped, leakage current from the Pch driver Tr may cause the output voltage to rise.
- 11. The average output voltage may vary due to the effects of output voltage ripple caused by the load current. Before use, test fully using the actual device.
- 12. If the C₁ capacitance or load current is large, the output voltage rise time will lengthen when the IC is started, and coil current overlay may occur during the interval until the output voltage reaches the set voltage (refer to the diagram below).



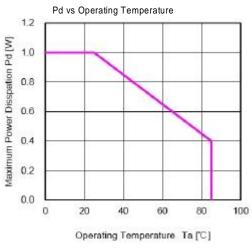
- 13. When the IC is started, the short-circuit protection function does not operate during the interval until the V our voltage reaches a value near the set voltage.
- 14. If the IC is started at a V_{IN} voltage that activates through mode, it is possible that the short-circuit protection function will not operate. Before use, test fully using the actual device.
- 15. If the load current is excessively large when the IC is started, it is possible that the Vout voltage will not rise to the set voltage. Before use, test fully using the actual device.

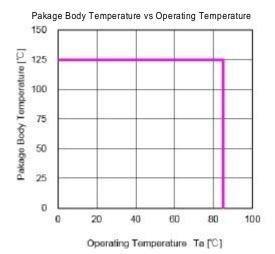
■NOTE ON USE (Continued)

- 16. In actual operation, the maximum on-time depends on the peripheral components, input voltage, and load current. Before use, test fully using the actual device
- 17. When the V_{IN} voltage is turned on and off continuously, excessive rush current may occur while the voltage is on. Before use, test fully using
- 18. When the V_{IN} voltage is high, the Pch driver may change from on to off before the coil current reaches the PFM switching current (I PFM), or before the maximum on-time elapses. Before use, test fully using the actual device.
- 19. When the IC change to the Through Mode at light load, the supply current of this IC can increase in some cases.
- 20. For temporary, transitional voltage drop or voltage rising phenomenon, the IC is liable to malfunction should the ratings be exceeded.
- 21. JIATAIMU places an importance on improving our products and their reliability.

 We request that users incorporate fail-safe designs and post-aging protection treatment when using JIATAIMU products in their systems.
- 22. The UVLO function can be activated when the UVLO hysteresis width gets to about 0mV and after several tens ms elapses at light loads. Before use, test fully using the actual device.
- 23. Please use within the power dissipation range below. power dissipation figure shown is PCB mounted.

Please also note that the power dissipation may changed by test conditions, the $\,$





the power loss of micro DC/DC according to the following formula:

power loss =
$$V_{OUT} \times I_{OUT} \times ((100/EFFI) - 1)$$
 (W)

Vour : Output Voltage (V)

IouT : Output Current (A)

EFFI : Conversion Efficiency (%)

Measurement Condition (Reference data)

Condition: Mount on a board Ambient: Natural convection Soldering: Lead (Pb) free

Board: Dimensions 40 x 40 mm (1600 mm₂ in one side)
Copper (Cu) traces occupy 50% of the board area

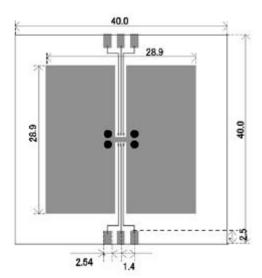
In top and back faces

Package heat-sink is tied to the copper traces

Material: Glass Epoxy (FR-4)

Thickness: 1.6mm

Through-hole: 4 x 0.8 Diameter



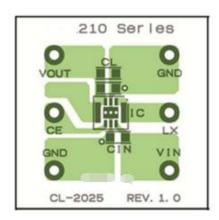
Evaluation Board (unit: mm)

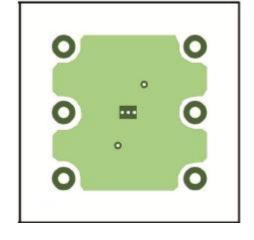
■NOTE ON USE (Continued)

•Instructions of pattern layouts

- 1. To suppress fluctuations in the V_{IN} potential, connect a bypass capacitor (C_{IN}) in the shortest path between the V_{IN} pin and ground pin.
- 2. Please mount each external component as close to the IC as possible.
- 3. Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
- 4. Make sure that the ground traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
- 5. Internal driver transistors bring on heat because of the transistor current and ON resistance of the driver transistors.

•Recommended Pattern Layout



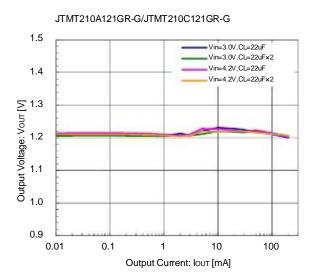


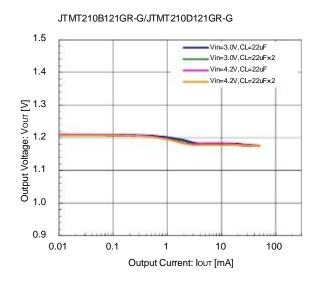
Top view

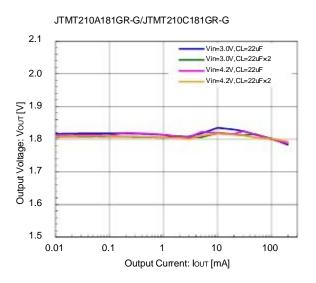
Back side top view

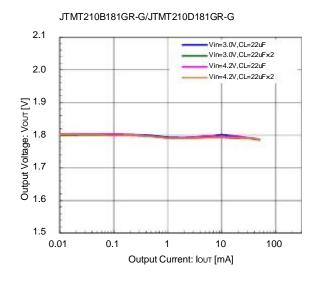
■TYPICAL PERFORMANCE CHARACTERISTICS

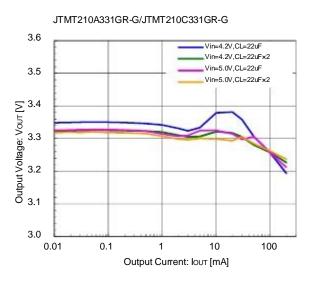
1) Output Voltage vs. Output Current

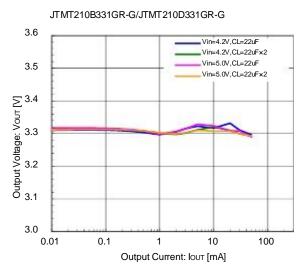






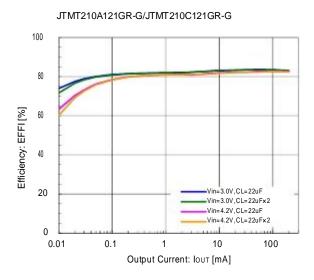


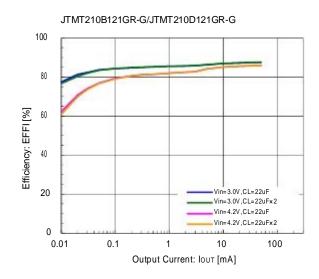


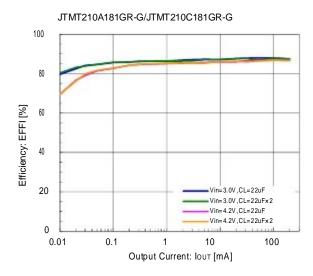


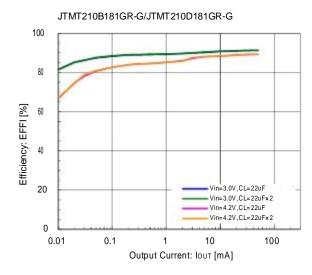
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

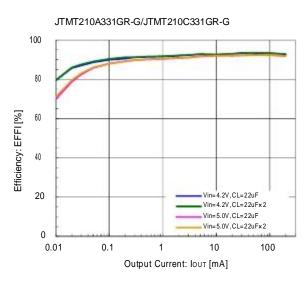
2) Efficiency vs. Output Current

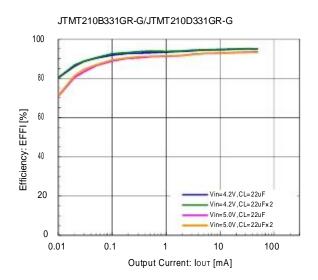






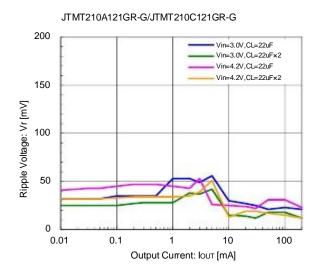


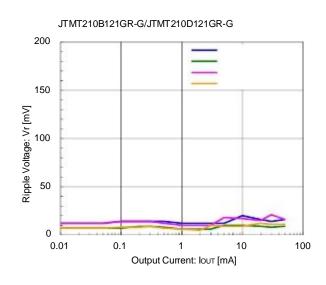


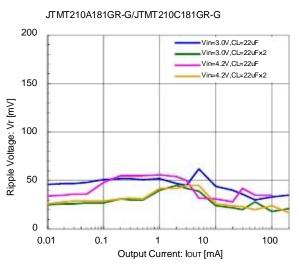


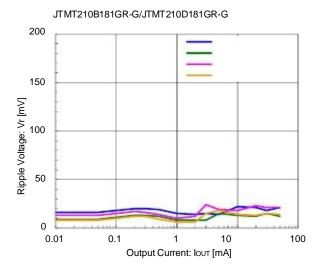
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

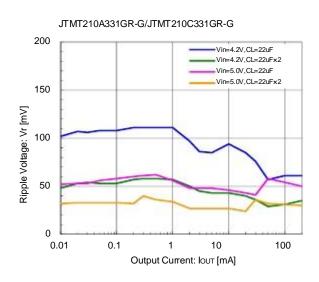
3) Ripple Voltage vs. Output Current

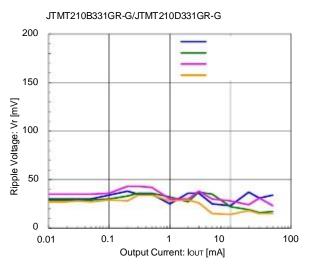






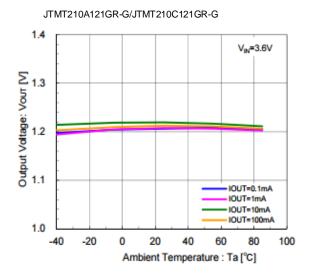


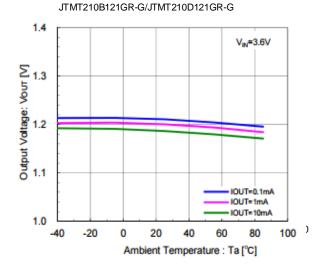


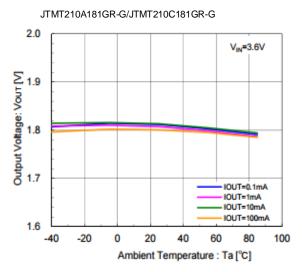


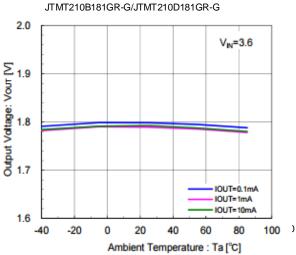
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

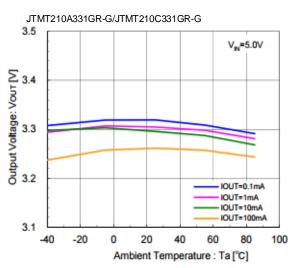
4) Ambient Temperature vs. Output Voltage

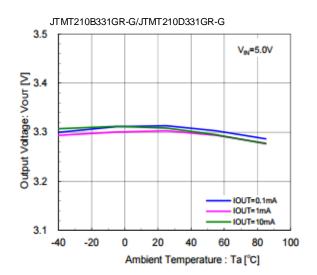








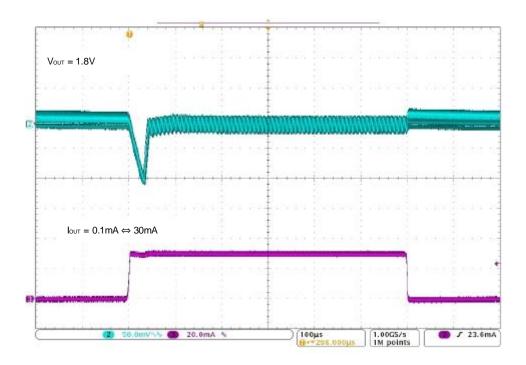




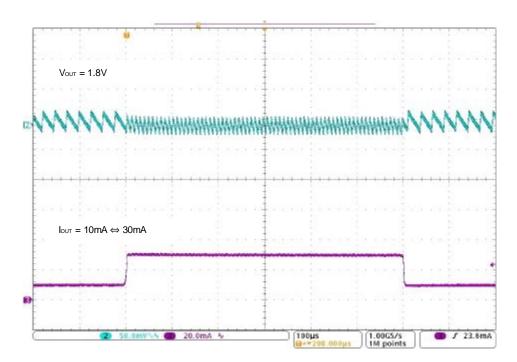
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

5) Load Transient Response

(1)JTMT210B181GR-G, V_{IN} =3.6V, V_{OUT} =1.8V / I_{OUT} =0.1mA \Leftrightarrow 30mA

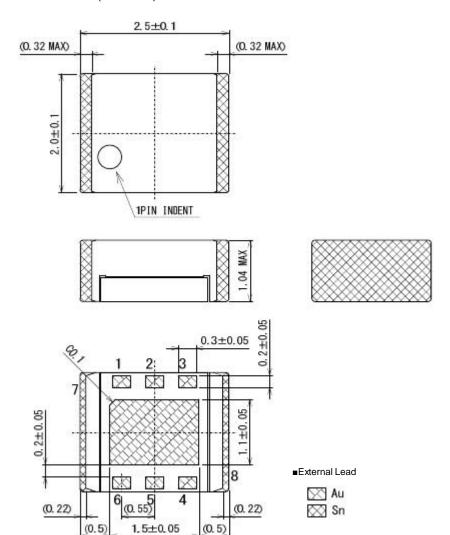


(2)JTMT210B181GR-G, V_{IN} =3.6V, V_{OUT} =1.8V / I_{OUT} =10mA \Leftrightarrow 30mA



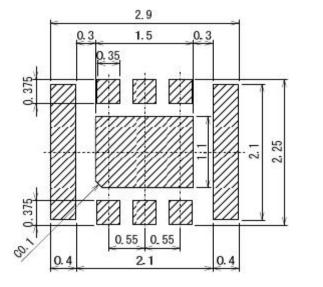
■PACKAGING INFORMATION

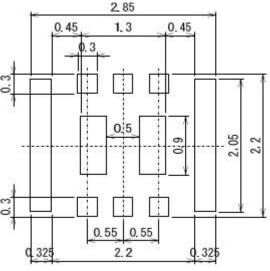
•CL-2025-02 (unit: mm)



•Reference Pattern Layout (unit: mm)

•Reference Metal Mask Design (unit: mm)

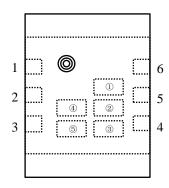




^{*} Implementation of CL-2025-02 is recommended within accuracy 0.05mm.

■MARKING RULE

●CL-2025-02



① represents products series

MARK	PRODUCT SERIES
0	JTMT210*****-G

② represents integer of the output voltage

MARK	TYPE	OUTPUT VOLTAGE(V)	PRODUCT SERIES
8		1.x	JTMT210A1****-G
9	Α	2.x	JTMT210A2****-G
Е		3.x	JTMT210A3****-G
F		4.x	JTMT210A4****-G
Н		1.x	JTMT210B1****-G
К	В	2.x	JTMT210B2****-G
L		3.x	JTMT210B3****-G
М		4.x	JTMT210B4****-G
N		1.x	JTMT210C1****-G
Р	С	2.x	JTMT210C2****-G
R		3.x	JTMT210C3****-G
S		4.x	JTMT210C4****-G
Т		1.x	JTMT210D1****-G
U	D	2.x	JTMT210D2****-G
V		3.x	JTMT210D3****-G
Х		4.x	JTMT210D4****-G

represents the decimal part of output voltage

_	represents the decimal part of output voltage			
-1	1 ₃ MARK	OUTPUT VOLTAGE(V)	PRODUCT SERIES	
	X.0	0	JTMT210**0***-G	
	X.05	A	JTMT210**A***-G	
	X.1	1	JTMT210**1***-G	
	X.15	В	JTMT210**B***-G	
	X.2	2	JTMT210**2***-G	
	X.25	С	JTMT210**C***-G	
	X.3	3	JTMT210**3***-G	
	X.35	D	JTMT210**D***-G	
	X.4	4	JTMT210**4***-G	
	X.45	E	JTMT210**E***-G	
	X.5	5	JTMT210**5***-G	
	X.55	F	JTMT210**F***-G	
	X.6	6	JTMT210**6***-G	
	X.65	Н	JTMT210**H***-G	
	X.7	7	JTMT210**7***-G	
	X.75	K	JTMT210**K***-G	
	X.8	8	JTMT210**8***-G	
	X.85	L	JTMT210**L***-G	
	X.9	9	JTMT210**9***-G	
	X.95	М	JTMT210**M***-G	

④,⑤ represents production lot number

01 \sim 09、0A \sim 0Z、11 \sim 9Z、A1 \sim A9、AA \sim AZ、B1 \sim ZZ in order.

(G, I, J, O, Q, W eJTMTuded)

Note: No character inversion used.

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