ETR28005-001a

### 1.5A Inductor Built-in Step-Down "micro DC/DC" Converters

☆GreenOperationCompatible

#### ■GENERAL DESCRIPTION

The JTMT213/JTMT214 series is a synchronous step-down micro DC/DC converter which integrates an inductor and a control IC in one tiny package (2.5mm×3.2mm, h=1.0mm). An internal coil simplifies the circuit and enables minimization of noise and other operational trouble due to the circuit wiring. A wide operating voltage range of 2.7V to 5.5V enables support for applications that require an internally fixed output voltage (0.8V to 3.6V). The JTMT213/JTMT214 series uses synchronous rectification at an operating frequency of 3.0MHz. The JTMT213/JTMT214 series uses HiSAT-COT(\*) synchronous rectification. HiSAT-COT+PWM control (JTMT213) or HiSAT-COT+automatic PWM/PFM switching control (JTMT214) can be selected. The series have a high speed soft-start as fast as 0.3ms in typical for quick turn-on. With the built-in UVLO (Under Voltage Lock Out) function, the internal P-channel driver transistor is forced OFF when input voltage becomes 2.0V or lower. When CE=Low, the integrated C<sub>L</sub> discharge function which enables the electric charge at the output capacitor C<sub>L</sub> to be discharged via the internal discharge switch located between the Lx and Vss pins. The power consumption will be less than 1.0μA.

#### ■APPLICATIONS

- Note Book PC
- ◆Tablet PC
- SSD
- Mobile phone
- Projector
- Printer

#### **■**FEATURES

 $\begin{array}{lll} \text{Input Voltage} & : 2.7 \text{V} \sim 5.5 \text{V} \\ \text{Output Voltage} & : 0.8 \text{V} \sim 3.6 \text{V} \\ \text{Oscillation Frequency} & : 3.0 \text{MHz} \\ \text{Output Current} & : 1.5 \text{A} \\ \end{array}$ 

Efficiency : 92% (V<sub>IN</sub>=5.0V, V<sub>OUT</sub>=3.3V/500mA)

Control Methods : HiSAT-COT

100% Duty Cycle

HiSAT-COT+PWM (JTMT213) HiSAT-COT+PWM/PFM (JTMT214)

Circuit Protection : Thermal Shut Down

Current Limit Circuit (Drop)
Short Circuit Protection (Latch)

Functions : Soft-start Circuit Built-in

UVLO

C∟ Discharge

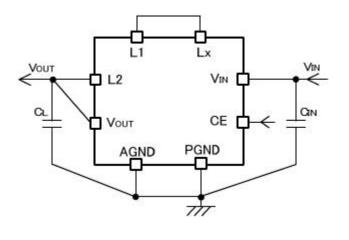
Output Capacitor : Low ESR Ceramic Capacitor

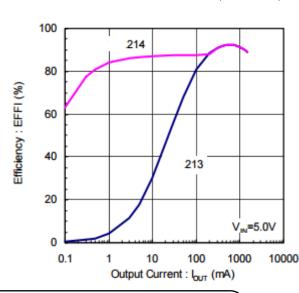
Operating Ambient Temperature :  $-40^{\circ}\text{C} \sim +85^{\circ}\text{C}$ Package : USP-9B01

Environmental Friendly : EU RoHS Compliant, Pb Free

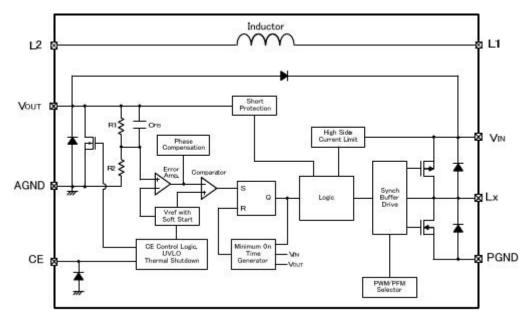
# ■TYPICAL APPLICATION CIRCUIT ■ TYPICAL PERFORMANCE CHARACTERISTICS

JTMT213/JTMT214 (Vout=3.3V)





# ■BLOCK DIAGRAM



<sup>\*</sup> The JTMT213 offers a fixed PWM control, a Control Logic of PWM/PFM Selector is fixed at "PWM" internally.

The JTMT214 control scheme is a fixed PWM/PFM automatic switching, a Control Logic of PWM/PFM Selector is fixed at "PWM/PFM automatic switching" internally.

Diodes inside the circuit are an ESD protection diode and a parasitic diode.

## **■PRODUCT CLASSIFICATION**

#### Ordering Information

JTMT213(1)2(3)4(5)6 PWM

JTMT214123456 PWM/PFM Automatic switching control

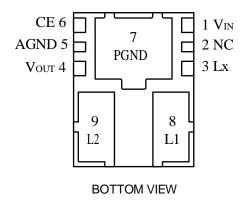
DESIGNATOR	ITEM	SYMBOL	DESCRIPTION
1)	Type	В	Refer to Selection Guide
23	Output Voltage	08~36	Output Voltage options e.g.)1.2V → ②=1, ③=2 1.25V → ②=1, ③=C  0.05V Increments: 0.05=A,0.15=B,0.25=C,0.35=D,0.45=E,0.55=F,0.65=H, 0.75=K,0.85=L,0.95=M
4	Oscillation Frequency	3	3.0MHz
56	Package (Order Unit)	DR	USP-9B01 (3,000/Reel)

The JTMT213/JTMT214 reels are shipped in a moisture-proof packing.

#### Selection Guide

TYPE	OUTPUTT VOLTAGE	C∟ AUTO- DISCHARGE	LATCH or SHORT PROTECTION	UVLO	CHIP ENABLE	CURRENT LIMIT	SOFT- START	THERMAL SHUTDOWN
В	Fixed	Yes	Yes	Yes	Yes	Yes	Fixed	Yes

### **■PIN CONFIGURATION**



### **■PIN ASSIGNMENT**

PIN NUMBER	PIN NAME	FUNCTIONS				
1	Vin	Power Input				
2	NC	No Connection				
3	Lx	Switching Output				
4	Vout	Fixed Output Voltage PIN				
5	AGND	Analog Ground				
6	CE	Chip Enable				
7	PGND	Power Ground				
8	L1	Inductor Electrodes				
9	L2	Inductor Electrodes				

## **■**FUNCTION TABLE

**CE PIN Function** 

PIN NAME	SIGNAL	STATUS
1	Low	Stand-by
2	High	Active

<sup>\*</sup> Please do not leave the CE pin open.

### ■ABSOLUTE MAXIMUM RATINGS

Ta=25°C

PARAMETER	SYMBOL	RATINGS	UNITS
V <sub>IN</sub> Pin Voltage	Vin	-0.3 ~ + 6.2	V
Lx Pin Voltage	VLx	- 0.3 ~ V <sub>IN</sub> + 0.3 or + 6.2 <sup>(*1)</sup>	V
Vουτ Pin Voltage	Vout	- 0.3 ~ Vin + 0.3 or + 4.0 (*2)	V
CE Pin Voltage	Vce	- 0.3 ~ + 6.2	V
Power Dissipation	Pd	1200	mW
Operating Ambient Temperature	Topr	- 40 ~ + 85	$^{\circ}$ C
Storage Temperature	T <sub>stg</sub>	- 55 ~ <b>+</b> 125	$^{\circ}$

All voltages are described based on the GND (AGND and PGND) pin.

The maximum value should be either V<sub>IN</sub>+0.3 or +6.2 in the lowest voltage

The maximum value should be either V<sub>IN</sub>+0.3 or +4.0 in the lowest voltage

## **■ELECTRICAL CHARACTERISTICS**

#### •JTMT213BxxxDR/JTMT214BxxxDR

Ta=25℃

PARAMETER	OVALDOL						
TAUGUNETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Output Voltage	Vouт	When connected to external conponets, lour =30mA	<e-1></e-1>	<e-2></e-2>	<e-3></e-3>	V	1)
Operating Voltage Range	Vin		2.7	-	5.5	V	1)
Maximum Output Current	loutmax	When connected to external components, $V_{\text{IN}} = < C\text{-}1>$	1500	-	-	mA	1)
UVLO Voltage	Vuvlo	Vout=0.6V, Voltage which Lx pin holding "L" level	1.35	2.00	2.68	V	3
Quiescent Current (JTMT214)	l <sub>q</sub>	Vout =Vout(E) x1.1V	-	25	40	μА	2
Quiescent Current (JTMT213)	<b>l</b> q	Vout =Vout(E) ×1.1V	-	400	825	μА	2
Stand-by Current	Іsтв	Vce =0V	-	0.0	1.0	μΑ	2
Minimum ON time	tonmin	When connected to external components, $V_{\text{IN}} = V_{\text{CE}} = < C-1> \text{, lou}_{\text{T}} = 1\text{mA}$	<e-5></e-5>	<e-6></e-6>	<e-7></e-7>	ns	1
Thermal Shutdown	T <sub>TSD</sub>		-	150	-	$^{\circ}$	1
Thermal shutdown Hysteresis	Тнуѕ		-	30	-	°C	1
LxSW "H" ON Resistance	$R_{LXH}$	Vоит=0.6V, lьx=100mA (°3)	-	0.14	0.28	Ω	4
LxSW "L" ON Resistance (*4)	RLXL	Vouт=Vouт(г) <b>×</b> 1.1V, Lx=100mA <sup>(*3)</sup>	-	0.10	0.20	Ω	4
LxSW "H" Leakage Current	I <sub>Leak</sub> н	VIN=5.5V, VCE=0V, VOUT=0V, VLX=0V	-	0.0	1.0	μА	(5)
LxSW "L" Leakage Current	LeakH	VIN=5.5V, VCE=0V, VOUT=0V, VLX=5.5V	-	0.0	30.0	μА	(5)
Current Limit (*5)	Ілімн	Vout=0.6V, ILx until Lx pin oscillates	2.5	3.0	4.5	Α	6
Output Voltage Temperature Characteristics	$\Delta V_{\text{OUT}}/$ ( $V_{\text{OUT}} \cdot \Delta  ext{topr}$ )	louτ=30mA -40°C ≦Topr≦85°C	-	±100	-	ppm/℃	1
CE"H" Voltage	V <sub>СЕН</sub>	Vouτ=0.6V, Applied voltage to Vcε, Voltage changes Lx to "H" level	1.4	-	Vin	V	3
CE"L" Voltage	Vcel	V <sub>FB</sub> =0.6V, Applied voltage to V <sub>CE</sub> , Voltage changes Lx to "L" level	AGND	-	0.3	V	3
CE"H" Current	СЕН	V <sub>IN</sub> =5.5V, V <sub>CE</sub> =5.5V, V <sub>OUT</sub> =0V	-0.1	-	0.1	μА	(5)
CE"L" Current	ICEL	VIN=5.5V, VCE=0V, VOUT=0V	-0.1	-	0.1	μΑ	(5)
Soft-start Time	tss	$V_{\text{CE}}=0V \rightarrow 5.0V$ , $V_{\text{OUT}}=V_{\text{OUT}(T)}V \times 0.9$ After "H" is fed to CE, the time by when clocks are generated at Lx pin.	0.10	0.30	0.50	ms	3
Short Protection Threshold Voltage	Vshort	Sweeping Vour, Vour voltage which Lx becomes "L" level	0.17	0.27	0.37	V	3
C∟ Discharge	Rochg	Vce=0V, Vout=4.0V	50	210	300	Ω	7
Inductance	L	Test Freq.=1MHz	-	0.5	-	μН	-
Inductor Rated Current	loc	ΔT=+40deg	-	1.8	-	А	-

Unless otherwise stated,  $V_{IN}$ =5.0V,  $V_{CE}$ =5.0V,  $V_{OUT(T)}$ =Nominal Voltage

#### NOTE:

<sup>(1)</sup> When the difference between the input and the output is small, 100% duty might come up and internal control circuits keep P-ch driver turning on even though the output current is not so large.

If current is further pulled from this state, output voltage will decrease because of P-ch driver ON resistance.

<sup>(°2)</sup> Including UVLO detect voltage, hysteresis operating voltage range for UVLO release voltage.

 $<sup>^{(&#</sup>x27;3)}$  R<sub>LXI</sub>=(V<sub>IN</sub> - Lx pin measurement voltage) / 100mA, R<sub>LXI</sub>=Lx pin measurement voltage / 100mA

 $<sup>\,^{^{({}^{\</sup>ast}{}^{4})}}$  Design value for the JTMT214 series

<sup>(°5)</sup> Current limit denotes the level of detection at peak of coil current.

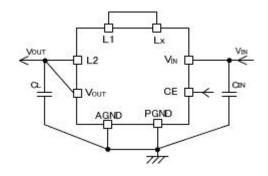
<sup>&</sup>quot;H"=V $_{\text{IN}}{\sim}$ V $_{\text{IN}}$  - 1.2V, "L"=- 0.1V ${\sim}$ + 0.1V

# **■**ELECTRICAL CHARACTERISTICS

#### ◆SPEC Table (Vout, tonmin)

NOMINAL		Vouт		i.e	ton	min	-	NOMINAL	Vouт			ton	lmin		
OUTPUT		<b>V</b> 001			fos	sc=3.0MH	z	OUTPUT				fo	sc=3.0MH	7	
VOLTAGE	<e-1></e-1>	<e-2></e-2>	<e-3></e-3>	<c-1></c-1>	<e-5></e-5>	<e-6></e-6>	<e-7></e-7>	VOLTAGE	<e-1></e-1>	<e-2></e-2>	<e-3></e-3>	<c-1></c-1>	<e-5></e-5>	<e-6></e-6>	<e-7></e-7>
V <sub>OUT(T)</sub>	MIN.	TYP.	MAX.	V <sub>IN</sub>	MIN.	TYP.	MAX.	<b>V</b> оит(т)	MIN.	TYP.	MAX.	VIN	MIN.	TYP.	MAX.
0.80	0.784	0.800	0.816	2.70	71	119	166	2.25	2.205	2.250	2.295	3.75	140	200	260
0.85	0.833	0.850	0.867	2.70	72	121	169	2.30	2.254	2.300	2.346	3.83	140	200	260
0.90	0.882	0.900	0.918	2.70	73	122	171	2.35	2.303	2.350	2.397	3.92	140	200	260
0.95	0.931	0.950	0.969	2.70	74	123	172	2.40	2.352	2.400	2.448	4.00	140	200	260
1.00	0.980	1.000	1.020	2.70	86	123	160	2.45	2.401	2.450	2.499	4.08	140	200	260
1.05	1.029	1.050	1.071	2.70	91	130	169	2.50	2.450	2.500	2.550	4.17	140	200	260
1.10	1.078	1.100	1.122	2.70	95	136	177	2.55	2.499	2.550	2.601	4.25	140	200	260
1.15	1.127	1.150	1.173	2.70	99	142	185	2.60	2.548	2.600	2.652	4.33	140	200	260
1.20	1.176	1.200	1.224	2.70	104	148	193	2.65	2.597	2.650	2.703	4.42	140	200	260
1.25	1.225	1.250	1.275	2.70	108	154	201	2.70	2.646	2.700	2.754	4.50	140	200	260
1.30	1.274	1.300	1.326	2.70	112	160	209	2.75	2.695	2.750	2.805	4.58	140	200	260
1.35	1.323	1.350	1.377	2.70	117	167	217	2.80	2.744	2.800	2.856	4.67	140	200	260
1.40	1.372	1.400	1.428	2.70	121	173	225	2.85	2.793	2.850	2.907	4.75	140	200	260
1.45	1.421	1.450	1.479	2.70	125	179	233	2.90	2.842	2.900	2.958	4.83	140	200	260
1.50	1.470	1.500	1.530	2.70	130	185	241	2.95	2.891	2.950	3.009	4.92	140	200	260
1.55	1.519	1.550	1.581	2.70	134	191	249	3.00	2.940	3.000	3.060	5.00	140	200	260
1.60	1.568	1.600	1.632	2.70	138	198	257	3.05	2.989	3.050	3.111	5.08	140	200	260
1.65	1.617	1.650	1.683	2.75	140	200	260	3.10	3.038	3.100	3.162	5.17	140	200	260
1.70	1.666	1.700	1.734	2.83	140	200	260	3.15	3.087	3.150	3.213	5.25	140	200	260
1.75	1.715	1.750	1.785	2.92	140	200	260	3.20	3.136	3.200	3.264	5.33	140	200	260
1.80	1.764	1.800	1.836	3.00	140	200	260	3.25	3.185	3.250	3.315	5.42	140	200	260
1.85	1.813	1.850	1.887	3.08	140	200	260	3.30	3.234	3.300	3.366	5.50	140	200	260
1.90	1.862	1.900	1.938	3.17	140	200	260	3.35	3.283	3.350	3.417	5.50	142	203	264
1.95	1.911	1.950	1.989	3.25	140	200	260	3.40	3.332	3.400	3.468	5.50	144	206	268
2.00	1.960	2.000	2.040	3.33	140	200	260	3.45	3.381	3.450	3.519	5.50	146	209	272
2.05	2.009	2.050	2.091	3.42	140	200	260	3.50	3.430	3.500	3.570	5.50	148	212	276
2.10	2.058	2.100	2.142	3.50	140	200	260	3.55	3.479	3.550	3.621	5.50	151	215	280
2.15	2.107	2.150	2.193	3.58	140	200	260	3.60	3.528	3.600	3.672	5.50	153	218	284
2.20	2.156	2.200	2.244	3.67	140	200	260		.1.					-	-

# **■**TYPICAL CIRCUIT



#### NOTE:

The integrated Inductor can be used only for this DC/DC converter. Please do not use this inductor for other reasons.

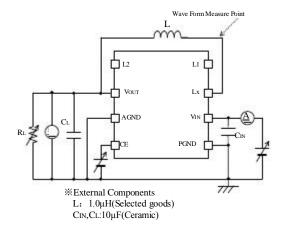
	VALUE	PRODUCT NUMBER
		LMK107BBJ106MALT(TaiyoYuden) (*1)
		LMK212ABJ106KG-T(TaiyoYuden) (*1)
CIN	10V/10μF	LMK212AB7106MG-T(TaiyoYuden) (*2)
CL		C1608JB1A106K (TDK) (*1)
		C2012JB1A106K (TDK) (*1)
		C2012X5R1A106K (TDK) (*1)
		C2012X7R1A106M (TDK) (*2)

#### NOTE:

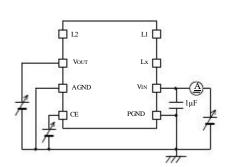
When under the condition of the voltage difference between input voltage and output voltage is low, please use 125  $^\circ\! {\rm C}$  product, which has small capacity drop.

# ■TEST CIRCUITS

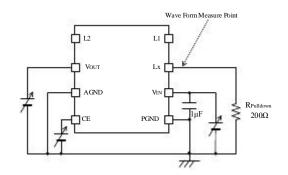
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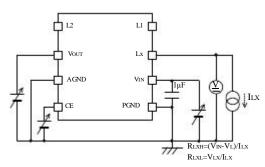
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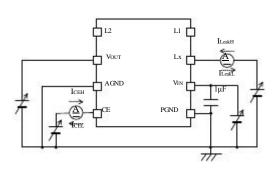
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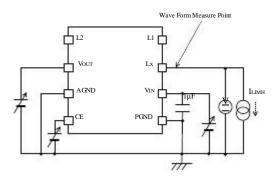
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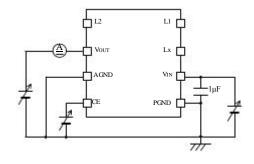
< Circuit No. 5 >



< Circuit No. 6 >

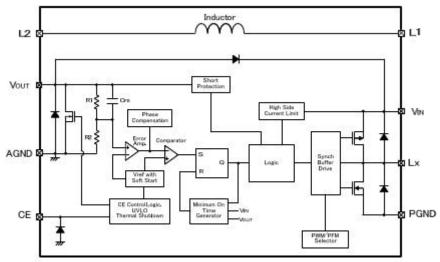


< Circuit No. 7 >



#### **■**OPERATIONAL EXPLANATION

The JTMT213/JTMT214 series consists of a reference voltage source, error amplifier, comparator, phase compensation, minimum on time generation circuit, output voltage adjustment resistors, P-channel MOS driver transistor, N-channel MOS switching transistor for the synchronous switch, current limiter circuit, UVLO circuit, thermal shutdown circuit, short protection circuit, PWM/PFM selection circuit and others. (See the BLOCK DIAGRAM below.)



<BLOCK DIAGRAM>

The method is HiSAT-COT (High Speed circuit Architecture for Transient with Constant On Time) control, which features on time control method and a fast transient response that also achieves low output voltage ripple.

The on time (ton) is determined by the input voltage and output voltage, and turns on the Pch MOS driver Tr. for a fixed time. During the off time (ton), the voltage that is fed back through R1 and R2 is compared to the reference voltage by the error amp, and the error amp output is phase compensated and sent to the comparator. The comparator compares this signal to the reference voltage, and if the signal is lower than the reference voltage, sets the SR latch. On time then resumes. By doing this, PWM operation takes place with the off time controlled to the optimum duty ratio and the output voltage is stabilized. The phase compensation circuit optimizes the frequency characteristics of the error amp, and generates a ramp wave similar to the ripple voltage that occurs in the output to modulate the output signal of the error amp. This enables a stable feedback system to be obtained even when a low ESR capacitor such as a ceramic capacitor is used, and a fast transient response and stabilization of the output voltage are achieved.

#### <Minimum on time generation circuit>

Generates an on time that depends on the input voltage and output voltage ( $t_{on}$ ). The on time is set as given by the equations below.  $t_{on}$  (ns) =  $V_{OUT}/V_{IN} \times 333$ 

#### <Switching frequency>

The switching frequency can be obtained from the on time (ton), which is determined by the input voltage and output voltage, as given by the equation below.

$$fosc(MHz) = Vout(V) / (Vin(V) \times ton(ns))$$

#### <100% duty cycle mode>

When the load current is heavy and the voltage difference between input voltage and output voltage is small, 100% duty cycle mode is activated and it keeps the Pch MOS driver Tr. turning on. 100% duty cycle mode attains a high output voltage stability and a high-speed response under all load conditions, from light to heavy, even in conditions where the dropout voltage is low.

#### <Error amp>

The error amp monitors the output voltage. The voltage divided by the internal R1 and R2 resistors is a feedback voltage for Error Amp. and compared to the reference voltage. The output voltage of the error amp becomes higher when the feedback voltage is higher than the reference voltage. The frequency characteristics of the error amp are optimized internally.

## ■OPERATIONAL EXPLANATION (Continued)

#### <Current Limit>

The current limiter circuit of the JTMT213/JTMT214 series monitors the current flowing through the P-channel MOS driver transistor connected to the Lx pin. When the driver current is greater than a specific level, the current limit function operates to turn off the pulses from the Lx pin at any given timing. When the over current state is eliminated, the IC resumes its normal operation.

#### <Reference voltage source, soft-start function>

The reference voltage forms a reference that is used to stabilize the output voltage of the IC. After chip enable of the IC, the reference voltage connected to the error amp increases linearly during the soft-start interval.

This allows the voltage divided by the internal R1 and R2 resistors and the reference voltage to be controlled in a balanced manner, and the output voltage rises in proportion to the rise in the reference voltage. This operation prevents rush input current and enables the output voltage to rise smoothly.

If the output voltage does not reach the set output voltage within the soft start time, such as when the load is heavy or a large capacity output capacitor is connected, the balancing of the voltage divided by the internal resistors R1 and R2 and the reference voltage is lost, however, the current restriction function activates to prevent an excessive increase of input current, enabling a smooth rise of the output voltage.

#### <PWM/PFM selection circuit>

PWM control is a continuous conduction mode, and operates at a stable switching frequency by means of an on time (ton) that is determined by the input voltage and output voltage regardless of the load.

PWM/PFM auto switching control is a discontinuous conduction mode at light loads, and lowers the switching frequency to reduce switching loss and improve efficiency.

The JTMT213 series is internally fixed to PWM control.

The JTMT214 series is internally fixed to PWM/PFM auto switching control.

#### <CE function>

Operation starts when "H" voltage is input into the CE pin. The IC can be put in the shutdown state by inputting "L" voltage into the CE pin. In the shutdown state, the supply current of the IC is  $0\mu$ A (TYP.), and the Pch MOS driver Tr. and Nch MOS switch Tr. for synchronous rectification turn off. The CE pin is a CMOS input and the sink current is  $0\mu$ A.

#### <UVLO>

When the  $V_{IN}$  voltage becomes 2.00V (TYP.) or lower, the P-ch MOS driver transistor output driver transistor is forced OFF to prevent false pulse output caused by unstable operation of the internal circuitry. When the  $V_{IN}$  pin voltage becomes 2.10V (TYP.) or higher, switching operation takes place. By releasing the UVLO function, the IC performs the soft start function to initiate output startup operation. The UVLO circuit does not cause a complete shutdown of the IC, but causes pulse output to be suspended; therefore, the internal circuitry remains in operation.

#### <Thermal Shutdown>

For protection against heat damage of the ICs, thermal shutdown function monitors chip temperature. The thermal shutdown circuit starts operating and the P-ch MOS driver and N-ch MOS driver transistor will be turned off when the chip's temperature reaches 150°C. When the temperature drops to 120°C (TYP.) or less after shutting of the current flow, the IC performs the soft-start function to initiate output startup operation.

#### <Short-circuit protection function>

The short-circuit protection circuit protects the device that is connected to this product and to the input/output in situations such as when the output is accidentally shorted to GND. The short-circuit protection circuit monitors the output voltage, and when the output voltage falls below the short-circuit protection threshold voltage, it turns off the Pch MOS driver Tr and latches it. Once in the latched state, operation is resumed by turning off the IC from the CE pin and then restarting, or by re-input into the V<sub>IN</sub> pin.

## ■OPERATIONAL EXPLANATION (Continued)

<CL High Speed Discharge>

The product can quickly discharge the electric charge at the output capacitor (CL) when a low signal to the CE pin which enables a whole IC circuit put into OFF state, is inputted via the N-ch MOS switch transistor located between the Vout pin and the GND pin. When the IC is disabled, electric charge at the output capacitor (CL) is quickly discharged so that it may avoid application malfunction.

Output Voltage Dischage characteristics

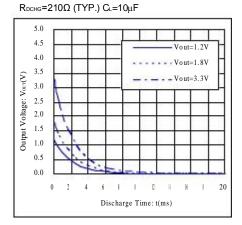
 $V=V_{OUT(T)} \times e^{-t/\tau}$  $t=\tau Ln (V_{OUT(T)} / V)$ 

V: Output voltage after discharge

Vouτ(τ): Output voltage t: Discharge time τ: CL×R<sub>DCHG</sub>

CL: Capacitance of Output capacitor

RDCHG: CL auto-discharge resistance,
but it depends on supply voltage.



#### ■NOTE ON USE

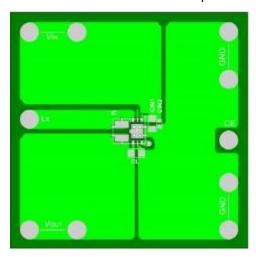
- 1. For the phenomenon of temporal and transitional voltage decrease or voltage increase, the IC may be damaged or deteriorated if IC is used beyond the absolute MAX. specifications.
- 2. Spike noise and ripple voltage arise in a switching regulator as with a DC/DC converter. These are greatly influenced by external component selection, such as the coil inductance, capacitance values, and board layout of external components. Once the design has been completed, verification with actual components should be done.
- 3. The DC/DC converter characteristics depend greatly on the externally connected components as well as on the characteristics of this IC, so refer to the specifications and standard circuit examples of each component when carefully considering which components to select. Be especially careful of the capacitor characteristics and use B characteristics (JIS standard) or X7R, X5R (EIA standard) ceramic capacitors.
- 4. Sufficiently reinforce the ground wiring. In particular, reinforce near the PGND and AGND pin as fluctuations of the ground phase due to the ground current during switching may cause the operation of the IC to become unstable.
- 5. Mount external components as close as possible to the IC. Keep the wiring short and thick to lower the wiring impedance.
- 6. A feature of HiSAT-COT control is that it controls the off time in order to control the duty, which varies due to the effects of power loss between the input (V<sub>IN</sub> pin) and output (V<sub>OUT</sub> pin) due to the load, and thus the switching frequency fluctuates. In addition, changes in the on time due to 100% duty cycle mode are allowed. For this reason, caution must be exercised as the characteristics of the switching frequency will vary depending on the external component characteristics, board layout, input voltage, output voltage, load current and other parameters.
- 7. Due to propagation delay inside the product, the on time generated by the minimum on time generation circuit may not be the same as the on time that is the ratio of the input voltage to the output voltage.
- 8. With regard to the current limiting value, the actual coil current may at times exceed the electrical characteristics due to propagation delay inside the product.
- 9. The CE pin is a CMOS input pin. Do not use with the pin open. If connecting to the input or ground, use a resistor up to 1 MΩ. To prevent malfunctioning of the device connected to this product or the input/output due to short circuiting between pins, it is recommended that a resistor be connected.
- 10. If the output voltage drops below the short circuit protection threshold voltage at the end of the soft start interval, operation will stop.
- 11. PWM/PFM auto switching control is a discontinuous conduction mode when the load is light, and in cases where the voltage difference between input and output is low or the coil inductance is higher than the value indicated in the standard circuit example, the coil current may reverse when the load is light, and thus pulse skipping will not be possible and light load efficiency will worsen.

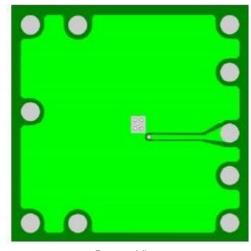
## ■NOTE ON USE (Continued)

- 12. When the voltage difference between input voltage and output voltage is low, the load stability feature may deteriorate.
- 13. JIATAIMU places an importance on improving our products and their reliability. We request that users incorporate fail-safe designs and post-aging protection treatment when using JIATAIMU products in their systems.
- 14. Instructions of pattern layouts
  - The operation may become unstable due to noise and/or phase lag from the output current when the wire impedance is high, please place the input capacitor( $C_{\mathbb{N}}$ ) and the output capacitor ( $C_{\mathbb{L}}$ ) as close to the IC as possible.
- (1) In order to stabilize V<sub>IN</sub> voltage level, we recommend that a by-pass capacitor (C<sub>IN</sub>) is connected as close as possible to the V<sub>IN</sub> pin, PGND pin and AGND pin.
- (2) Please mount each external component as close to the IC as possible.
- (3) Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
- (4) Make sure that the GND traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
- (5) This series' internal driver transistors bring on heat because of the output current and ON resistance of P-channel and N-channel MOS driver transistors. Please consider the countermeasures against heat if necessary.

#### <Reference pattern layout>

C<sub>IN</sub>2: Optional



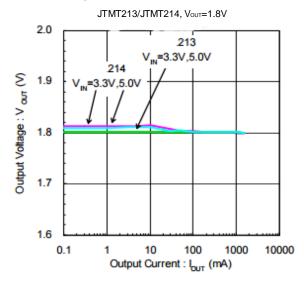


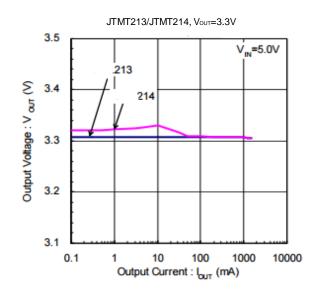
<Top View>

<Bottom View>

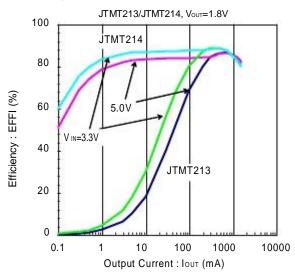
### **■**TYPICAL PERFORMANCE CHARACTERISTICS

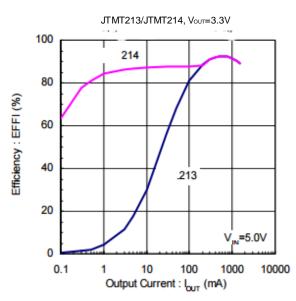
#### (1) Output Voltage vs. Output Current



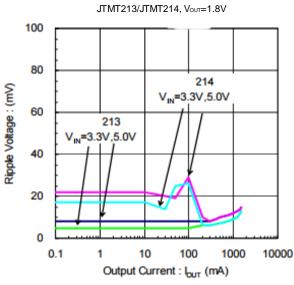


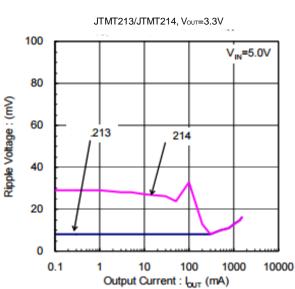
#### (2) Efficiency vs. Output Current





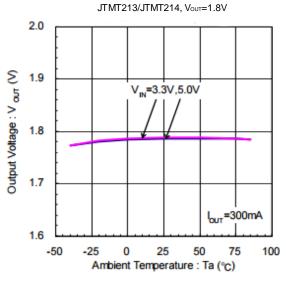
#### (3) Ripple Voltage vs. Output Current



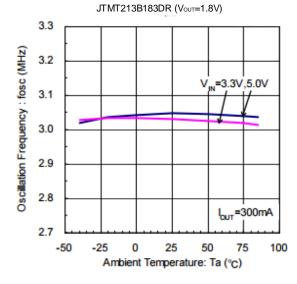


# ■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

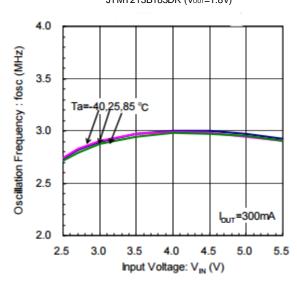
#### (4) Output Voltage vs. Ambient Temperature

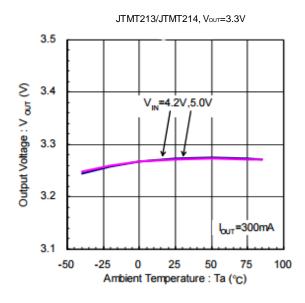


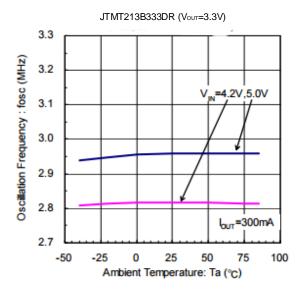
## (5) Oscillation Frequency vs. Ambient Temperature

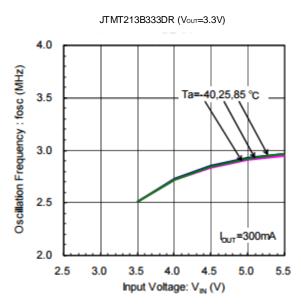


#### (6) Oscillation Frequency vs. Input Voltage JTMT213B183DR (Vout=1.8V)





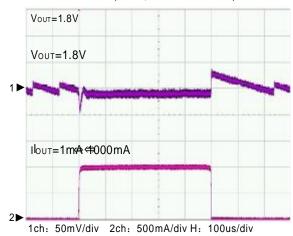


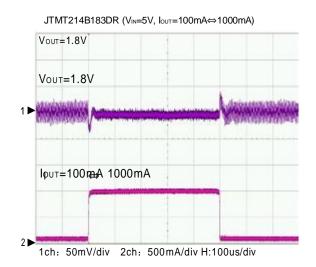


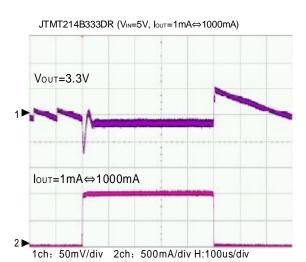
# ■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

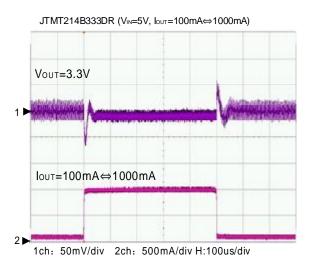
#### (7) Load Transient Response

JTMT214B183DR (V<sub>IN</sub>=5V, louт=1mA⇔1000mA)



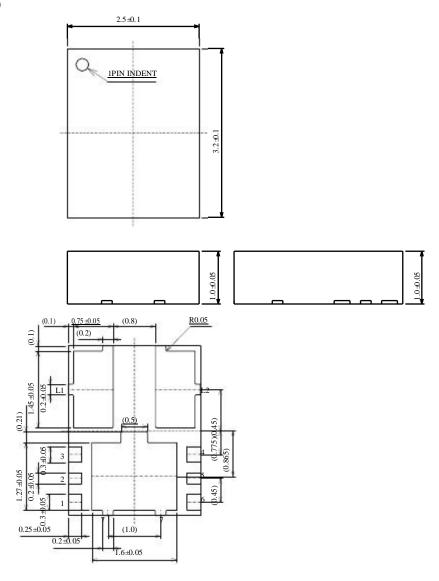






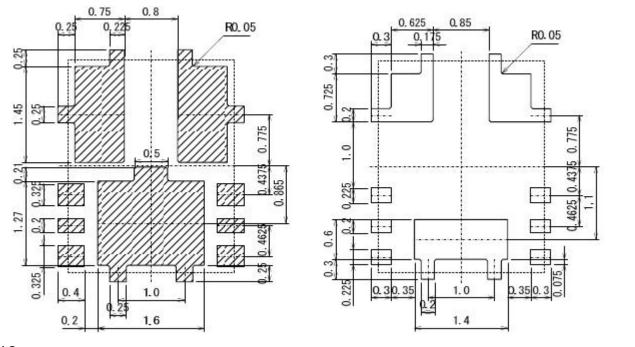
# **■PACKAGING INFORMATION**

•USP-9B01 (unit:mm)



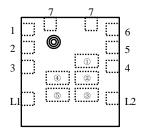
•USP-9B01 Reference Pattern Layout (unit: mm)

•USP-9B01 Reference Metal Mask Design (unit: mm)



## ■MARKING RULE

USP-9B01



#### ① represents products series

MARK	PRODUCT SERIES					
Α	JTMT213*****					
В	JTMT214*****					

#### ② represents integer of the reference voltage

OUTPUT VOLTAGE (V)	MARK	PRODUCT SERIES
0.X	А	JTMT21**0****
1.X	В	JTMT21**1****
2.X	С	JTMT21**2****
3.X	D	JTMT21**3****

#### ③ represents decimal number of the reference voltage

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OUTPUT VOLTAGE (V)	MARK	PRODUCT SERIES	OUTPUT VOLTAGE (V)	MARK	PRODUCT SERIES					
X.0	0	JTMT21***0***	X.05	Α	JTMT21***A***					
X.1	1	JTMT21***1***	X.15	В	JTMT21***B***					
X.2	2	JTMT21***2***	X.25	С	JTMT21***C***					
X.3	3	JTMT21***3***	X.35	D	JTMT21***D***					
X.4	4	JTMT21***4***	X.45	Е	JTMT21***E***					
X.5	5	JTMT21***5***	X.55	F	JTMT21***F***					
X.6	6	JTMT21***6***	X.65	Ι	JTMT21***H***					
X.7	7	JTMT21***7***	X.75	K	JTMT21***K***					
X.8	8	JTMT21***8***	X.85	L	JTMT21***L***					
X.9	9	JTMT21***9***	X.95	М	JTMT21***M***					

45 represents production lot number

01 $\sim$ 09, 0A $\sim$ 0Z, 11 $\sim$ 9Z, A1 $\sim$ A9, AA $\sim$ AZ, B1 $\sim$ ZZ in order.

(G, I, J, O, Q, W eJTMTuded)

<sup>\*</sup> No character inversion used.

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